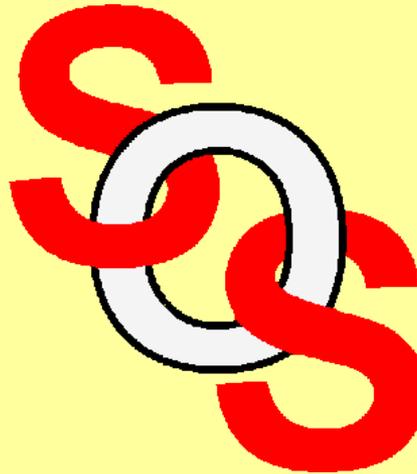


SPACE MAPPING BASED DEVICE MODELING AND CIRCUIT OPTIMIZATION

J.W. Bandler, M.H. Bakr, J.E. Rayas-Sánchez, M.A. Ismail and Q.S. Cheng

Simulation Optimization Systems Research Laboratory and
Department of Electrical and Computer Engineering
McMaster University, Hamilton, Canada L8S 4K1

bandler@mcmaster.ca
www.sos.mcmaster.ca



presented at

WORKSHOP ON OPTIMUM AND GLOBAL ELECTROMAGNETIC MODELLING USING
HYBRID TECHNIQUES FROM ANALYSIS TO OPTIMIZATION

30th European Microwave Conference 2000, Paris, France, October 6, 2000



SPACE MAPPING BASED DEVICE MODELING AND CIRCUIT OPTIMIZATION

J.W. Bandler, M.H. Bakr, J.E. Rayas-Sánchez, M.A. Ismail and Q.S. Cheng

bandler@mcmaster.ca

www.sos.mcmaster.ca

Abstract

Electromagnetics (EM) based device modeling and circuit optimization through Space Mapping (SM) technologies are reviewed. The SM concept continues to promise important benefits in the next generation of design optimization methodologies. Artificial Neural Networks can be incorporated into the SM optimization strategies. Aggressive Space Mapping (ASM) optimization closely follows the traditional experience and intuition of designers, while being rigorously grounded mathematically. Current progress in the development of suitable algorithms and software engines are presented. The SM concept addresses the contradictory challenge of exploitation of device models for CAD that are both accurate and fast.



Outline

a comprehensive Generalized Space Mapping (GSM) tableau approach (*Bandler et al., 1999*) to engineering device modeling exploiting Frequency Space Mapping (FSM) (*Bandler et al., 1995*) and the Multiple Space Mapping (MSM) (*Bandler et al., 1998*) is reviewed

a Neural Space Mapping (NSM) optimization approach exploiting our SM-based neuromodeling techniques is presented (*Bakr et al., 2000*)

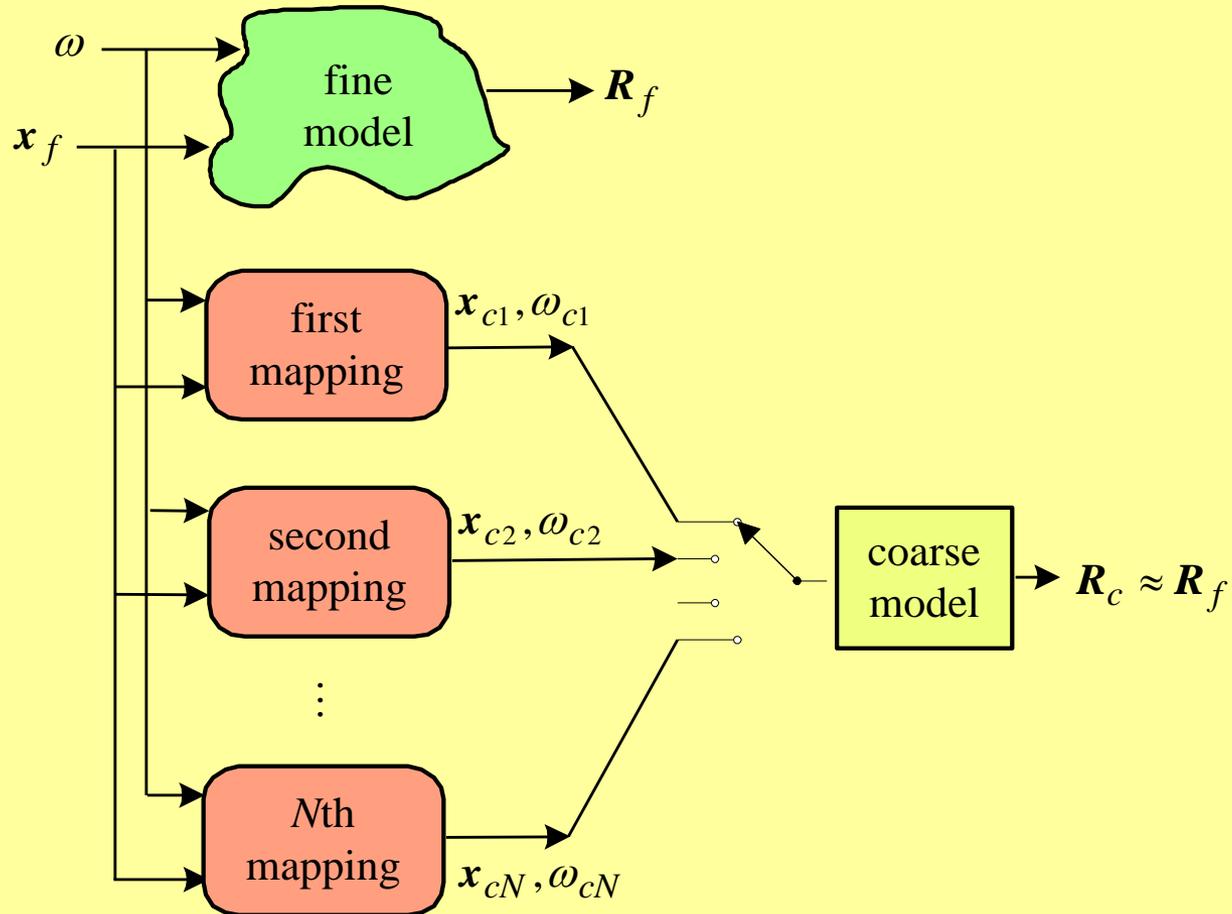
new work on Space Mapping optimization exploiting surrogate models is described (*Bakr et al., 2000*)

a state-of-the-art engineering optimization system including the latest Space Mapping technology, the SMX system, is described (*Bandler et al., 2000*)



Multiple Space Mapping (MSM) Concept

MSM for Frequency Intervals (MSMFI)





Mathematical Formulation for GSM

(Bandler et al., 1999)

the k th mapping targeting the sub-response or the response \mathbf{R} in the k th frequency sub-range is given by

$$(\mathbf{x}_{ck}, \omega_{ck}) = \mathbf{P}_k(\mathbf{x}_f, \omega)$$

or, in matrix form, assuming a linear mapping

$$\begin{bmatrix} \mathbf{x}_{ck} \\ \omega_{ck} \end{bmatrix} = \begin{bmatrix} \mathbf{c}_k \\ \delta_k \end{bmatrix} + \begin{bmatrix} \mathbf{B}_k & s_k \\ \mathbf{t}_k^T & \sigma_k \end{bmatrix} \begin{bmatrix} \mathbf{x}_f \\ \omega \end{bmatrix}$$

the mapping parameters $\{\mathbf{c}_k, \mathbf{B}_k, s_k, \mathbf{t}_k, \sigma_k, \delta_k\}$ can be evaluated, directly or indirectly, by solving the optimization problem

$$\min_{\mathbf{c}_k, \mathbf{B}_k, s_k, \mathbf{t}_k, \sigma_k, \delta_k} \left\| \begin{bmatrix} \mathbf{e}_{k1}^T & \mathbf{e}_{k2}^T & \cdots & \mathbf{e}_{km}^T \end{bmatrix}^T \right\|$$

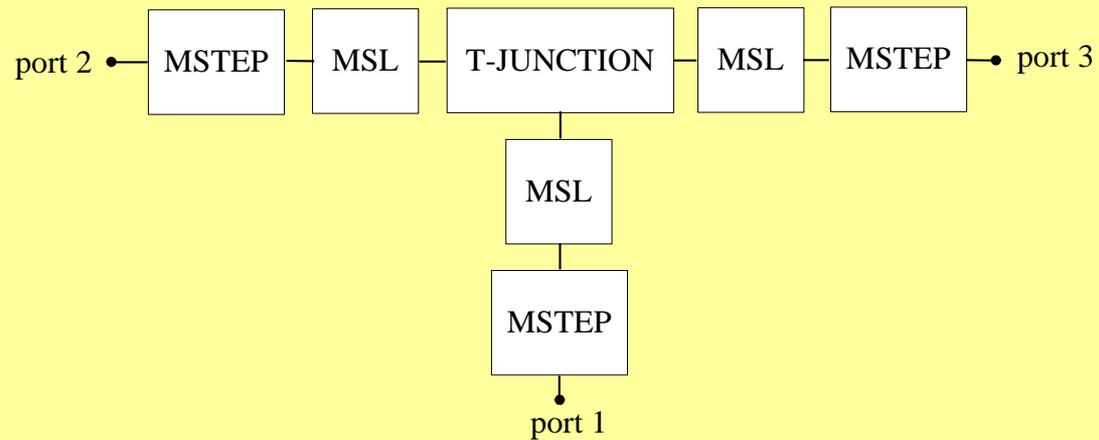
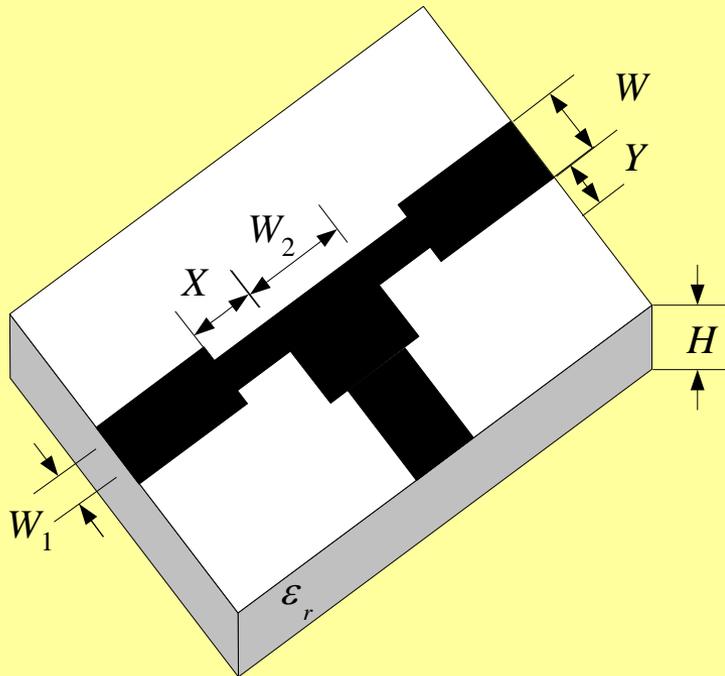
where m is the number of base points selected in the fine model space and \mathbf{e}_{kj} is an error vector given by

$$\mathbf{e}_{kj} = \mathbf{R}_f(\mathbf{x}_f^{(j)}, \omega) - \mathbf{R}_c(\mathbf{x}_{ck}^{(j)}, \omega_{ck}), \quad j = 1, 2, \dots, m$$



Microstrip Shaped T-Junction

the fine and coarse models





Microstrip Shaped T-Junction

the region of interest

$$15 \text{ mil} \leq H \leq 25 \text{ mil}$$

$$5 \text{ mil} \leq X \leq 15 \text{ mil}$$

$$5 \text{ mil} \leq Y \leq 15 \text{ mil}$$

$$8 \leq \varepsilon_r \leq 10$$

the frequency range is 2 GHz to 20 GHz with a step of 2 GHz

the number of base points is 9 and the number of test points is 50

the width W of the input lines is determined in terms of H and so that the characteristic impedance of the input lines is 50 ohm

the width W_1 is taken as 1/3 of the width W

the width W_2 is obtained so that the characteristic impedance of the microstrip line after the step connected to port 2 is twice that of the microstrip line after the step connected to port 1



Microstrip Shaped T-Junction

MSM for Frequency Intervals (MSMFI) was developed to enhance the accuracy of the T-Junction coarse model

the total frequency range was divided into two intervals: 2-16 GHz and 16-20 GHz

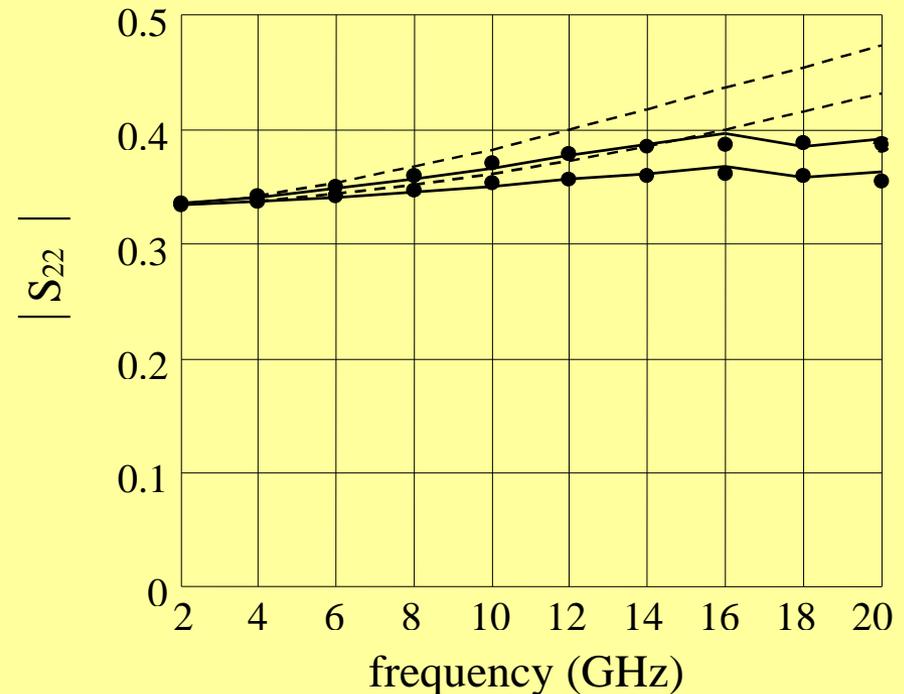
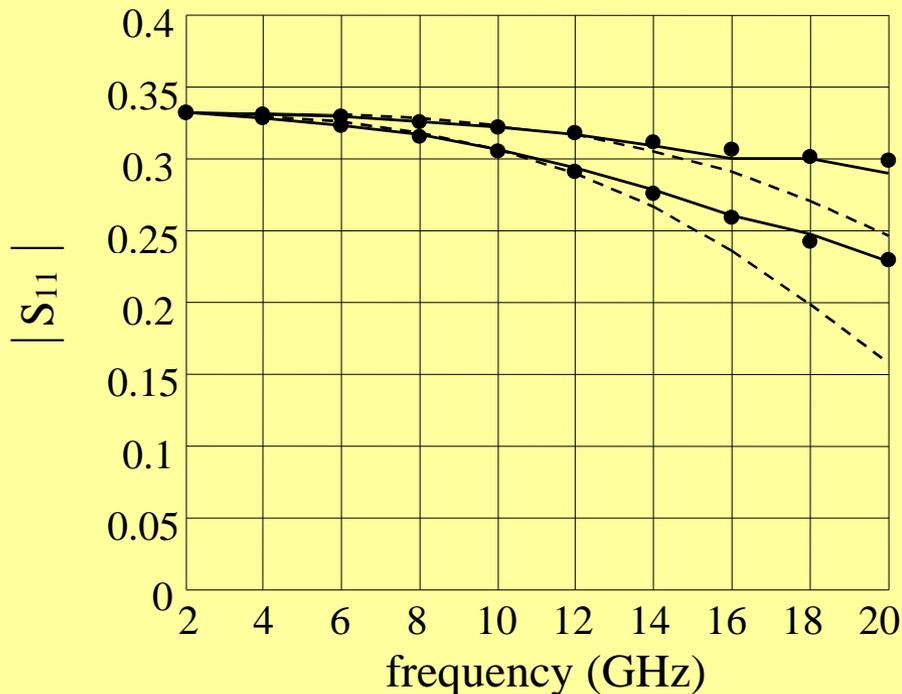
the mapping parameters are

	2 GHz to 16 GHz	16 GHz to 20 GHz
B	$\begin{bmatrix} 1.04 & 0.07 & 0.01 & 0.08 & -0.06 & 0.00 & 0.22 \\ 0.00 & 0.89 & 0.00 & -0.07 & -0.20 & 0.06 & -0.03 \\ -0.00 & 0.07 & 0.99 & 0.04 & -0.12 & 0.01 & -0.06 \\ -0.04 & 0.00 & -0.01 & 0.97 & 0.10 & -0.06 & -0.27 \\ 0.01 & 0.04 & 0.00 & 0.03 & 0.99 & -0.05 & -0.03 \\ -0.13 & -0.05 & -0.04 & -0.16 & 0.12 & 0.99 & 0.62 \\ -0.08 & 0.12 & -0.03 & 0.00 & -0.07 & 0.03 & 0.83 \end{bmatrix}$	$\begin{bmatrix} 0.99 & 0.02 & -0.00 & 0.01 & -0.09 & -0.01 & 0.13 \\ 0.05 & 0.85 & 0.01 & -0.07 & -0.28 & 0.01 & -0.01 \\ -0.06 & 0.15 & 0.98 & 0.04 & -0.25 & 0.00 & 0.02 \\ -0.10 & -0.06 & -0.03 & 0.88 & 0.13 & -0.09 & -0.27 \\ 0.08 & 0.04 & 0.03 & 0.11 & 1.07 & -0.04 & -0.12 \\ -0.14 & -0.02 & -0.05 & -0.15 & 0.23 & 1.03 & 0.51 \\ -0.13 & 0.22 & -0.04 & 0.02 & -0.07 & 0.03 & 0.87 \end{bmatrix}$
c	$[0.02 \quad 0.01 \quad -0.01 \quad -0.03 \quad -0.01 \quad 0.07 \quad -0.03]^T$	$[0.01 \quad 0.01 \quad -0.01 \quad -0.03 \quad -0.01 \quad 0.05 \quad -0.03]^T$
s	$[-0.01 \quad 0.09 \quad -0.10 \quad -0.02 \quad 0.00 \quad -0.02 \quad -0.20]^T$	$[0.00 \quad 0.01 \quad -0.01 \quad 0.00 \quad 0.00 \quad 0.00 \quad -0.02]^T$
t	$\mathbf{0}$	$[0.01 \quad 0.00 \quad -0.02 \quad 0.00 \quad 0.00 \quad 0.00 \quad 0.00]^T$
σ	0.851	0.957
δ	-0.003	0.008



Microstrip Shaped T-Junction

the responses of the shaped T-Junction at two test points in the region of interest by Sonnet's *em* (\bullet), by the coarse model (---) and by the enhanced coarse model (—)





Microstrip Shaped T-Junction

the enhanced coarse model for the shaped T-Junction can be utilized in optimization

the optimization variables are X and Y

the other parameters are kept fixed ($W = 24$ mil, $H = 25$ mil and $\epsilon_r = 9.9$)

the design specifications are

$$|S_{11}| \leq 1/3, \quad |S_{22}| \leq 1/3$$

in the frequency range 2 GHz to 16 GHz

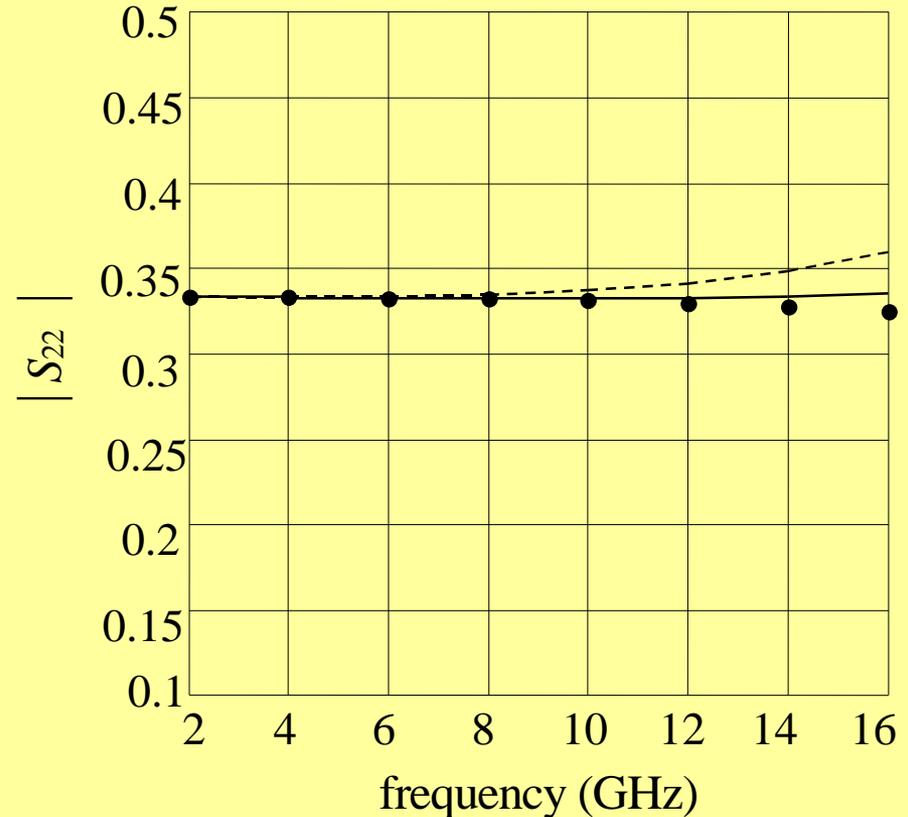
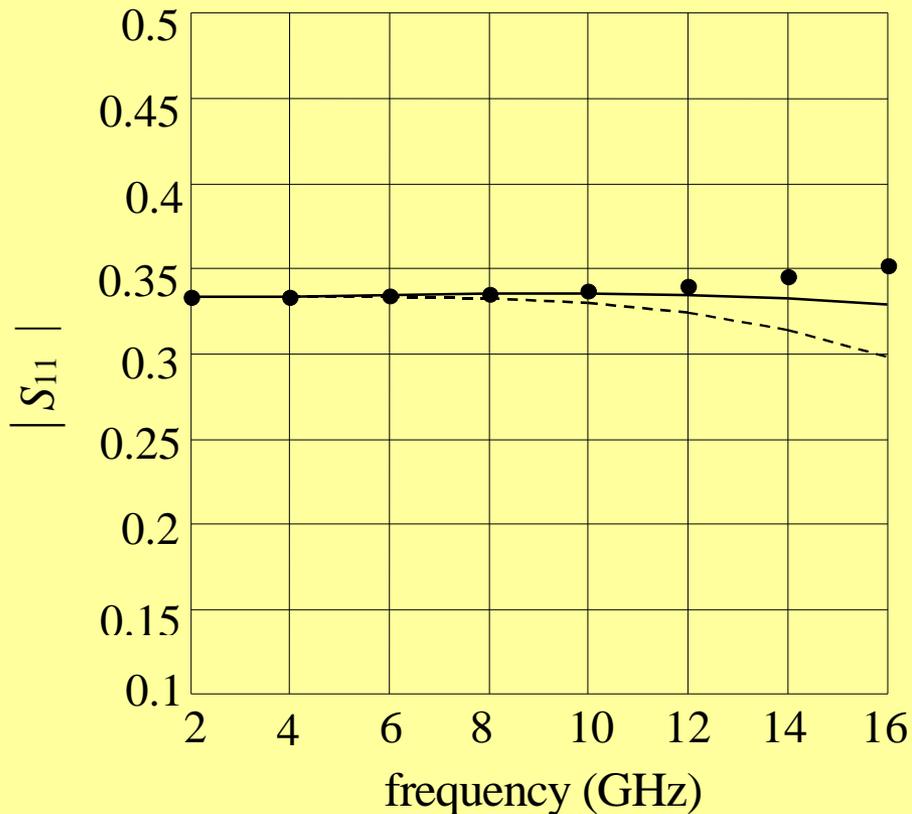
the minimax optimizer in OSA90/hope reached the solution

$$X = 2.1 \text{ mil and } Y = 21.1 \text{ mil}$$



Microstrip Shaped T-Junction

responses of the optimal shaped T-Junction by Sonnet's *em* (\bullet), by the coarse model (---) and by the enhanced coarse model (—)





Neural Space Mapping (NSM) Optimization

(Bakr et al., 2000)

exploits the SM-based neuromodeling techniques

(Bandler et al., 1999)

coarse models are used as sources of knowledge that reduce the amount of learning data and improve the generalization and extrapolation performance

NSM requires a reduced set of upfront learning base points

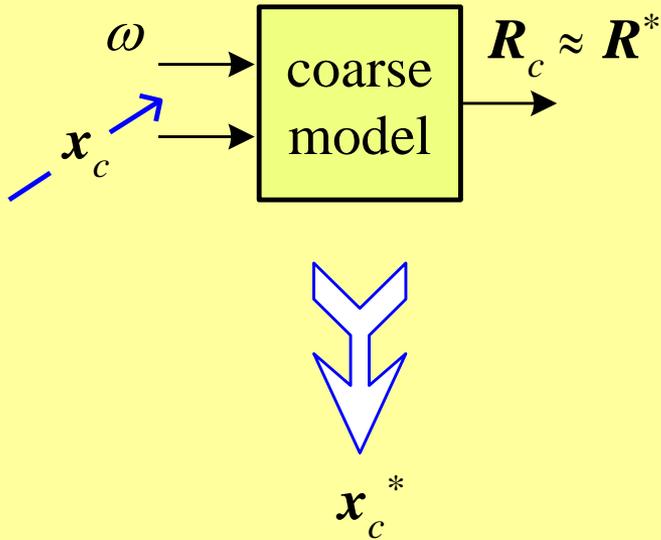
the initial learning base points are selected through sensitivity analysis using the coarse model

neuromappings are developed iteratively: their generalization performance is controlled by gradually increasing their complexity starting with a 3-layer perceptron with 0 hidden neurons

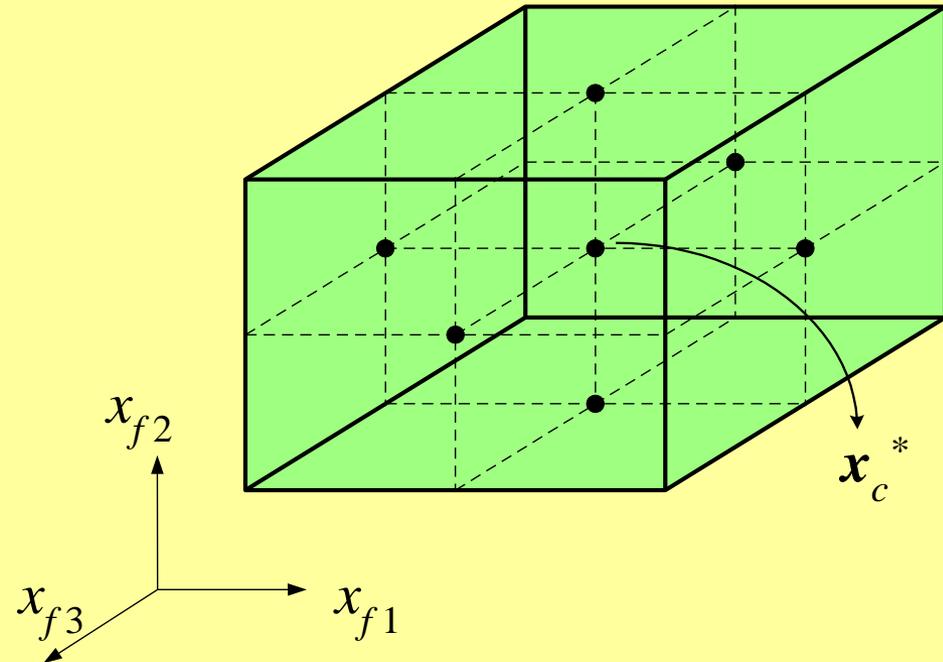


Neural Space Mapping (NSM) Optimization Concept

step 1



step 2

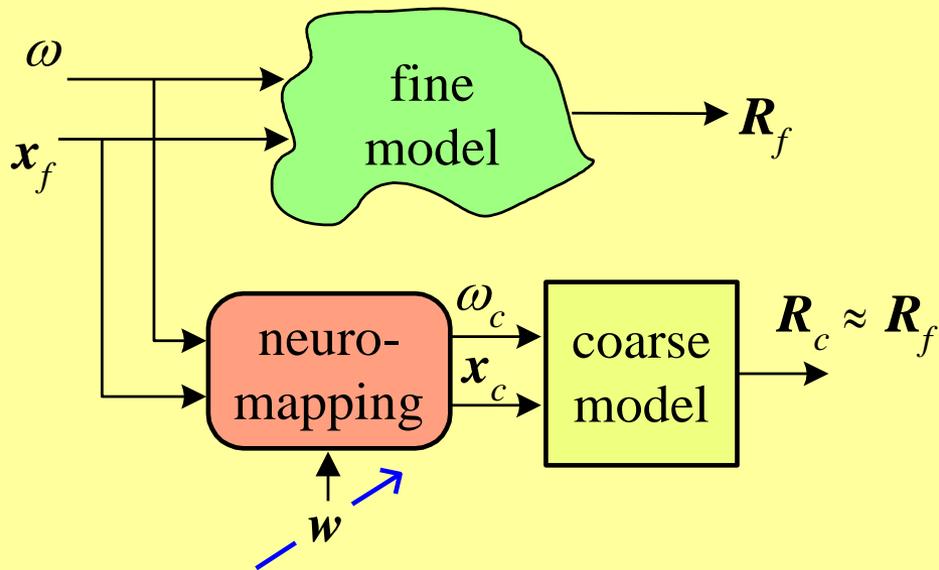


($2n + 1$ learning base points for a microwave circuit with n design parameters)

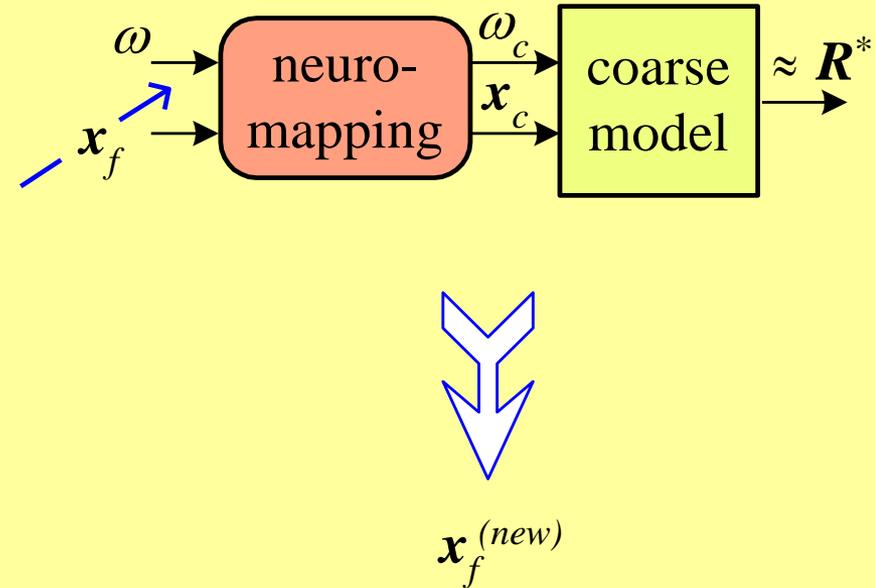


Neural Space Mapping (NSM) Optimization Concept (continued)

step 3

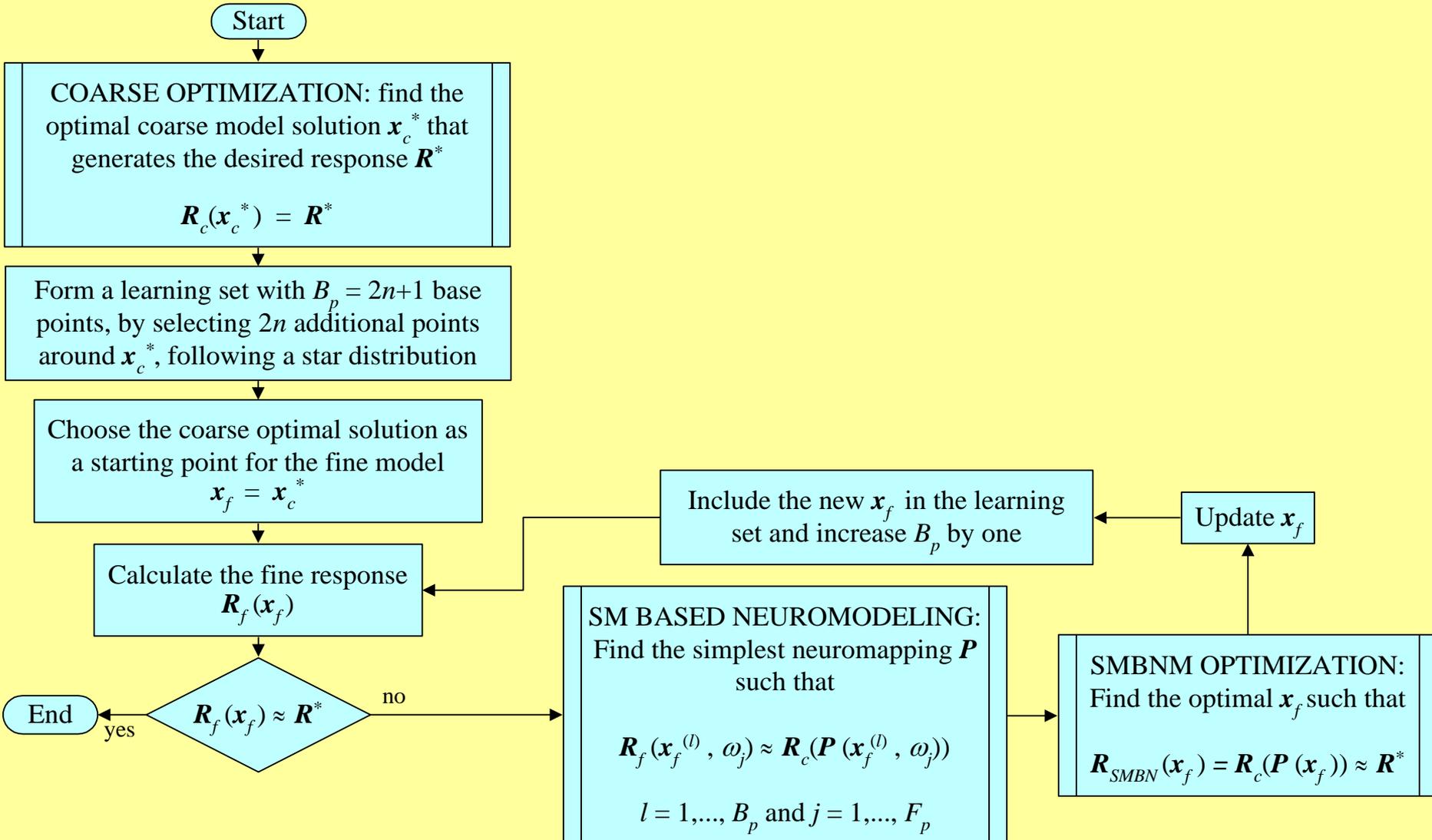


step 4





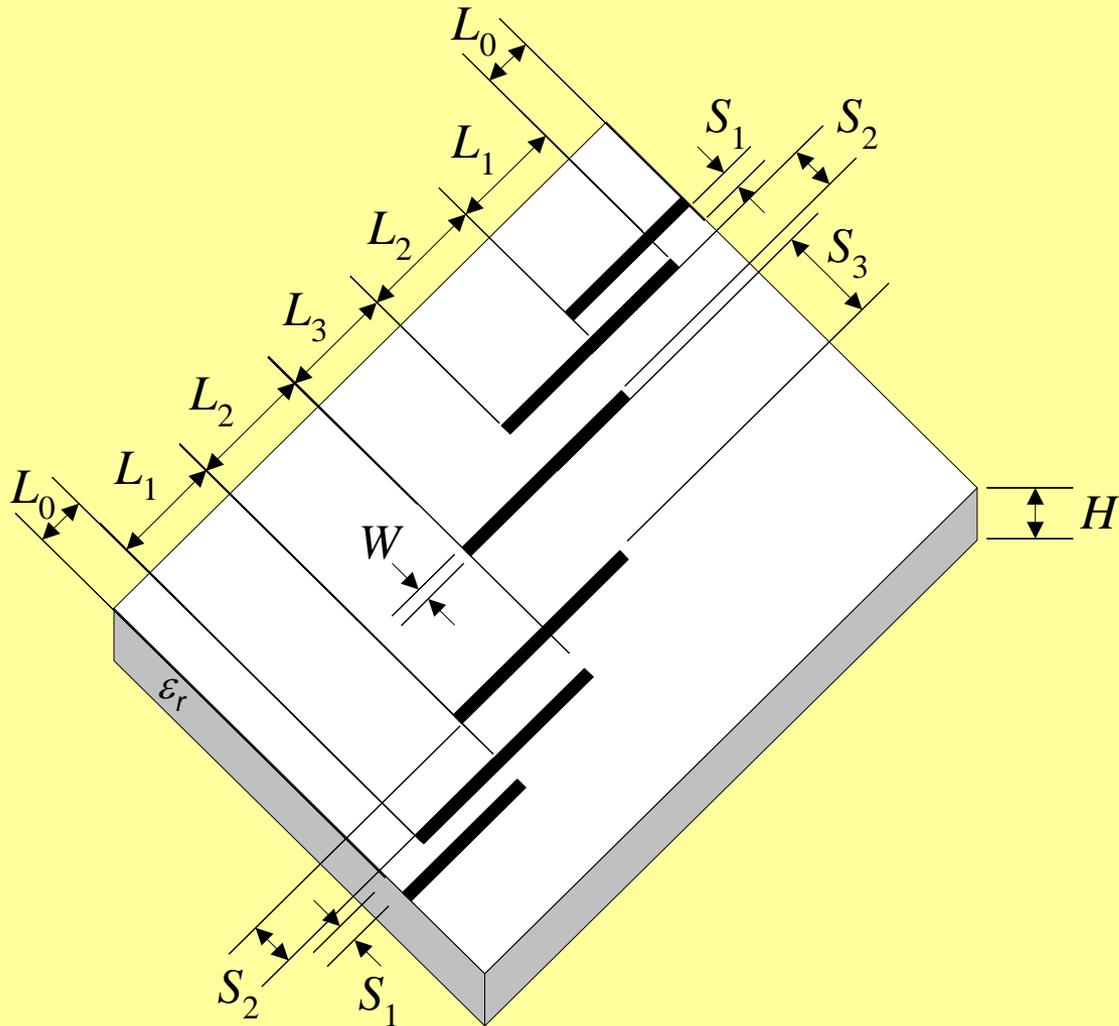
Neural Space Mapping (NSM) Optimization Algorithm





HTS Quarter-Wave Parallel Coupled-Line Microstrip Filter

(Westinghouse, 1993)



we take $L_0 = 50$ mil, $H = 20$ mil,
 $W = 7$ mil, $\epsilon_r = 23.425$, loss
tangent = 3×10^{-5} ; the
metalization is considered
lossless

the design parameters are
 $\mathbf{x}_f = [L_1 \ L_2 \ L_3 \ S_1 \ S_2 \ S_3]^T$



NSM Optimization of the HTS Microstrip Filter

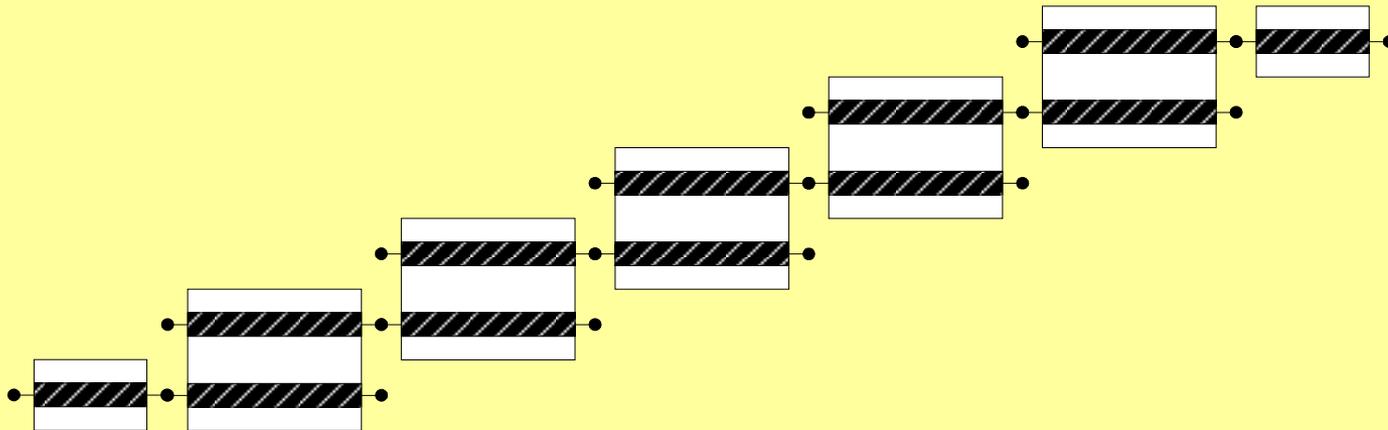
specifications

$$|S_{21}| \geq 0.95 \text{ for } 4.008 \text{ GHz} \leq f \leq 4.058 \text{ GHz}$$

$$|S_{21}| \leq 0.05 \text{ for } f \leq 3.967 \text{ GHz and } f \geq 4.099 \text{ GHz}$$

“fine” model: Sonnet’s *em*TM with high resolution grid

“coarse” model: OSA90/hopeTM built-in models of open circuits, microstrip lines and coupled microstrip lines



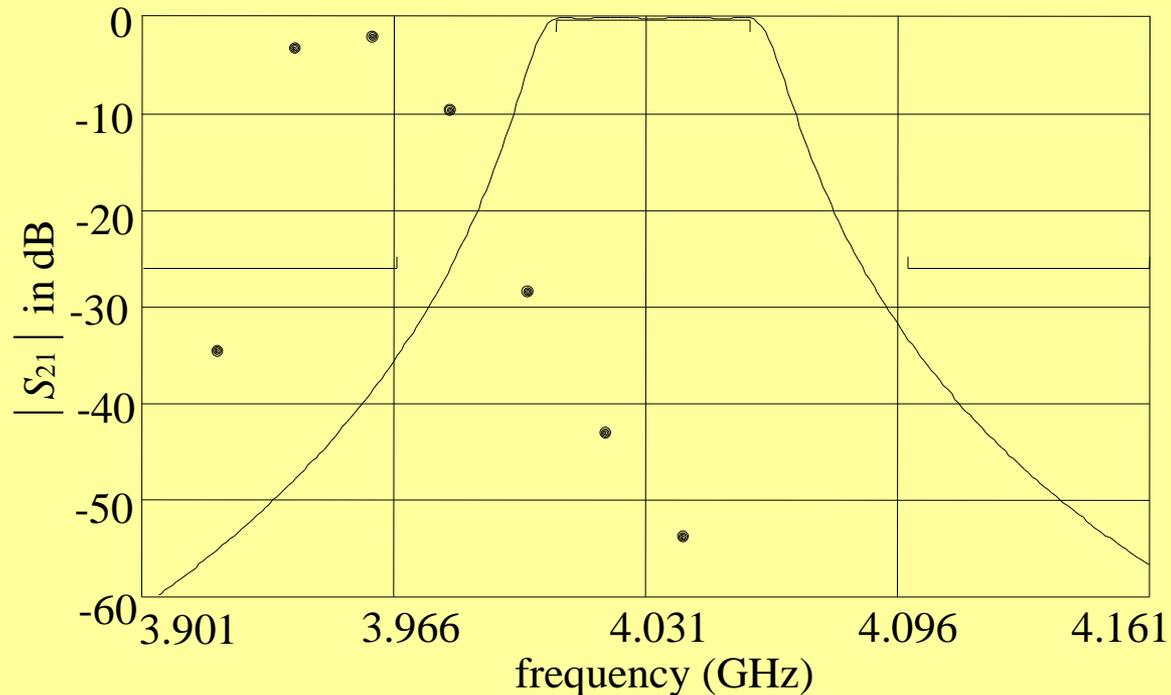


NSM Optimization of the HTS Filter (continued)

coarse and fine model responses at the optimal coarse solution

$$\mathbf{x}_c^* = [188.33 \ 197.98 \ 188.58 \ 21.97 \ 99.12 \ 111.67]^T \text{ (mils)}$$

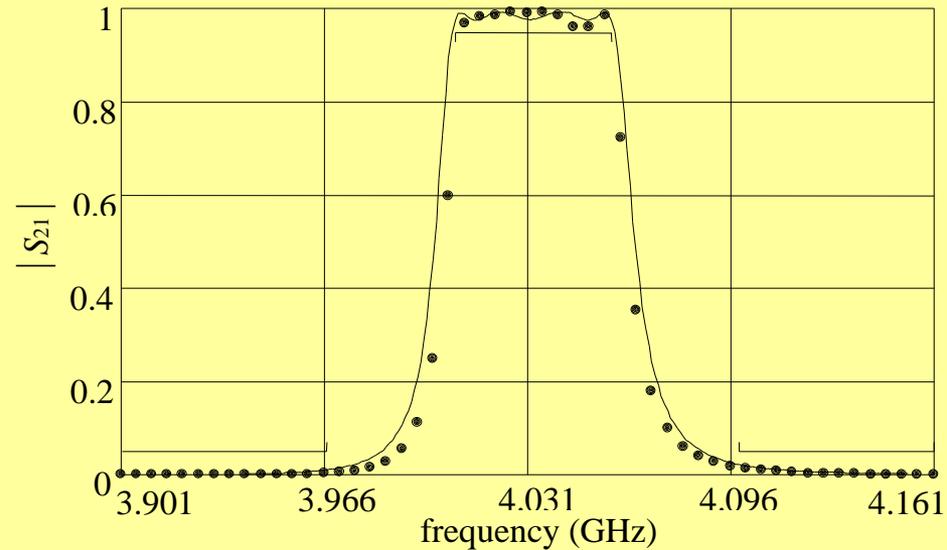
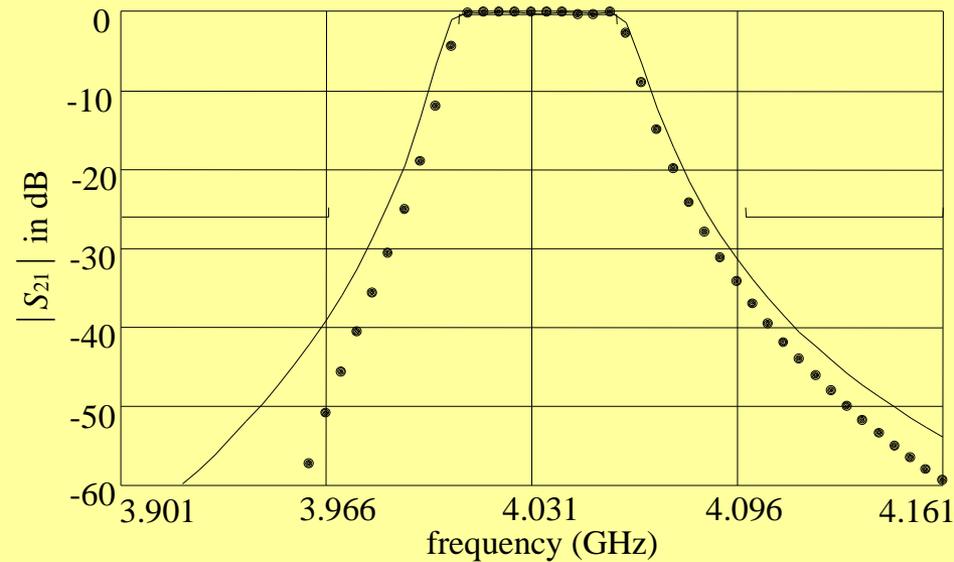
OSA90/hope™ (—) and *em*™ (●)





NSM Optimization of the HTS Filter (continued)

em^{TM} (●) and FPSM 7-5-3 (—) model responses at the NSM solution using a fine frequency sweep





Space Mapping Optimization Exploiting Surrogates

(Bakr et al., 2000)

a powerful new Space Mapping (SM) optimization algorithm has been developed

it draws upon recent developments in both surrogate model-based optimization and modeling of microwave devices

SM optimization is formulated as a general optimization problem of a surrogate model

this model is a convex combination of a mapped coarse model and a linearized fine model

it exploits, in a novel way, a linear frequency-sensitive mapping

during the optimization iterates, the coarse and fine models are simulated at different sets of frequencies.

this approach is shown to be especially powerful if a significant response shift exists



The Surrogate Model

our surrogate model is a convex combination of a mapped coarse model and a linearized fine model

the i th iteration surrogate model is

$$\mathbf{R}_s^{(i)}(\mathbf{x}_f) = \lambda^{(i)} \mathbf{R}_m^{(i)}(\mathbf{x}_f) + (1 - \lambda^{(i)}) (\mathbf{R}_f(\mathbf{x}_f^{(i)}) + \mathbf{J}_f^{(i)} \Delta \mathbf{x}_f), \quad \lambda^{(i)} \in [0, 1]$$

the mapped coarse model utilizes the frequency-sensitive mapping

$$\mathbf{R}_f(\mathbf{x}_f, \omega_j) \approx \mathbf{R}_m^{(i)}(\mathbf{x}_f, \omega_j) = \mathbf{R}_c(\mathbf{P}^{(i)}(\mathbf{x}_f, \omega_j), \mathbf{P}_\omega^{(i)}(\mathbf{x}_f, \omega_j))$$

where

$$\begin{bmatrix} \mathbf{P}^{(i)}(\mathbf{x}_f, \omega_j) \\ \mathbf{P}_\omega^{(i)}(\mathbf{x}_f, \omega_j) \end{bmatrix} = \begin{bmatrix} \mathbf{B}^{(i)} & \mathbf{s}^{(i)} \\ \mathbf{t}^{(i)T} & \sigma^{(i)} \end{bmatrix} \begin{bmatrix} \Delta \mathbf{x}_f \\ \omega_j \end{bmatrix} + \begin{bmatrix} \mathbf{c}^{(i)} \\ \gamma^{(i)} \end{bmatrix}$$

the parameters $\mathbf{B}^{(i)} \in \mathfrak{R}^{n \times n}$, $\mathbf{s}^{(i)} \in \mathfrak{R}^{n \times 1}$, $\mathbf{t}^{(i)} \in \mathfrak{R}^{n \times 1}$, $\mathbf{c}^{(i)} \in \mathfrak{R}^{n \times 1}$, $\sigma^{(i)} \in \mathfrak{R}^{1 \times 1}$ and $\gamma^{(i)} \in \mathfrak{R}^{1 \times 1}$ are obtained such that the mapped coarse model approximates the fine model over a given set of fine model points $V^{(i)}$ and frequencies ω



The Surrogate Model (continued)

the mapping parameters are obtained through the optimization process

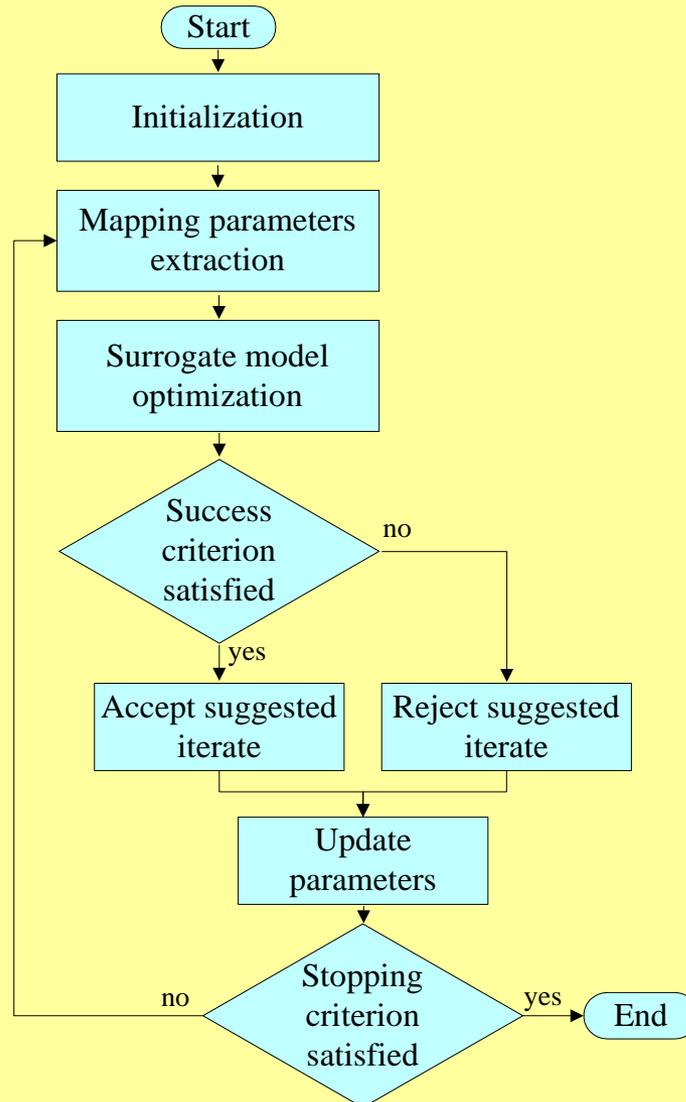
$$[\mathbf{B}^{(i)}, \mathbf{s}^{(i)}, \mathbf{t}^{(i)}, \sigma^{(i)}, \mathbf{c}^{(i)}, \gamma^{(i)}] = \arg \left\{ \min_{\mathbf{B}, \mathbf{s}, \mathbf{t}, \sigma, \mathbf{c}, \gamma} \left\| \begin{bmatrix} \mathbf{e}_1^T & \mathbf{e}_2^T & \cdots & \mathbf{e}_{N_p}^T \end{bmatrix}^T \right\| \right\}$$

where

$$\mathbf{e}_k = \mathbf{R}_m^{(i)}(\mathbf{x}_f^{(k)}) - \mathbf{R}_f(\mathbf{x}_f^{(k)}) \quad \forall \mathbf{x}_f^{(k)} \in V^{(i)}$$



The Algorithm Flowchart





The SMX System

(Bandler et al., 2000)

SMX is a new generation engineering optimization system

currently it provides the following optimization capabilities

- minimax

- Huber

- Space Mapping using Surrogate Models *(Bakr et al., 2000)*

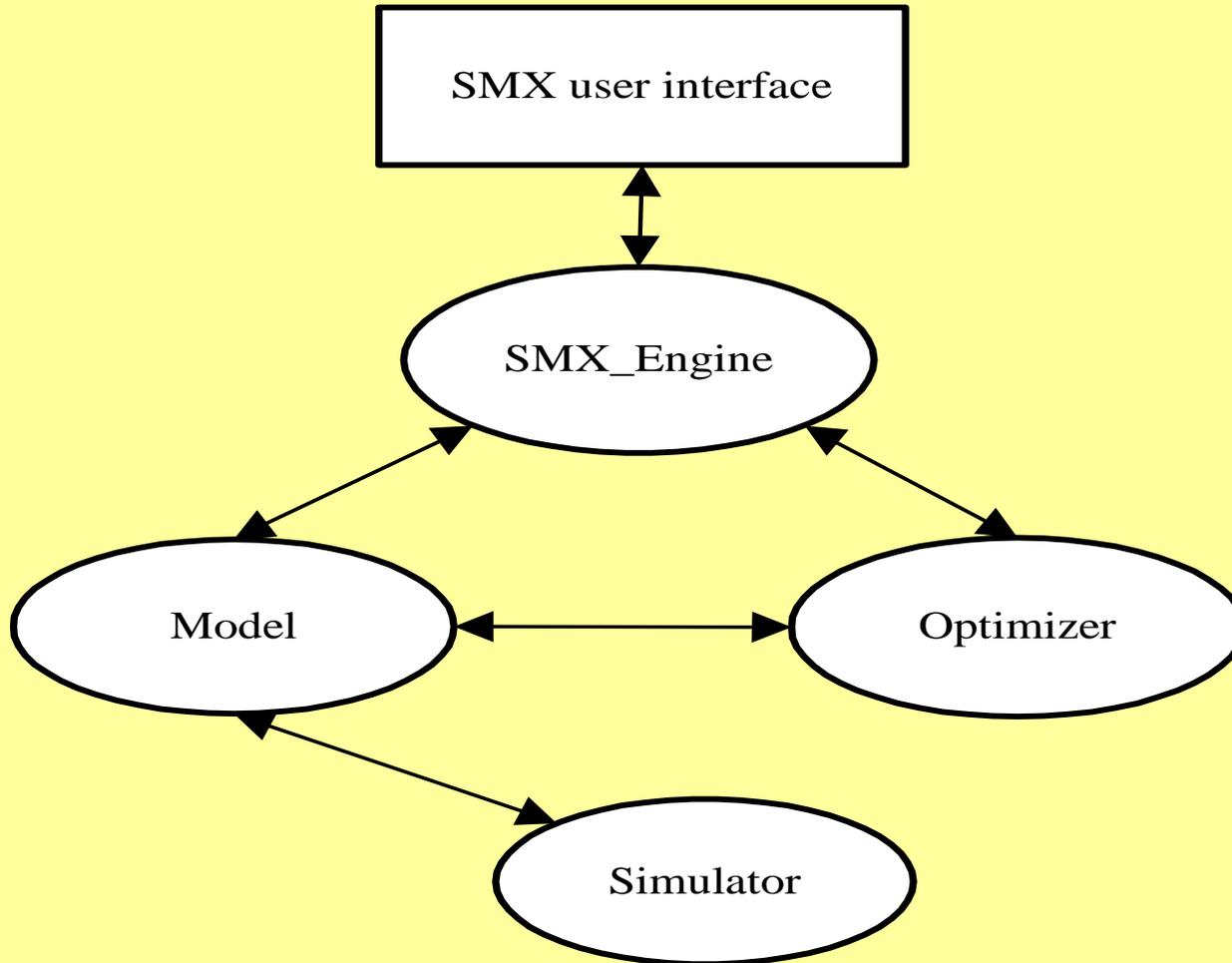
currently it can be interfaced to

- OSA90

- user supplied executable programs



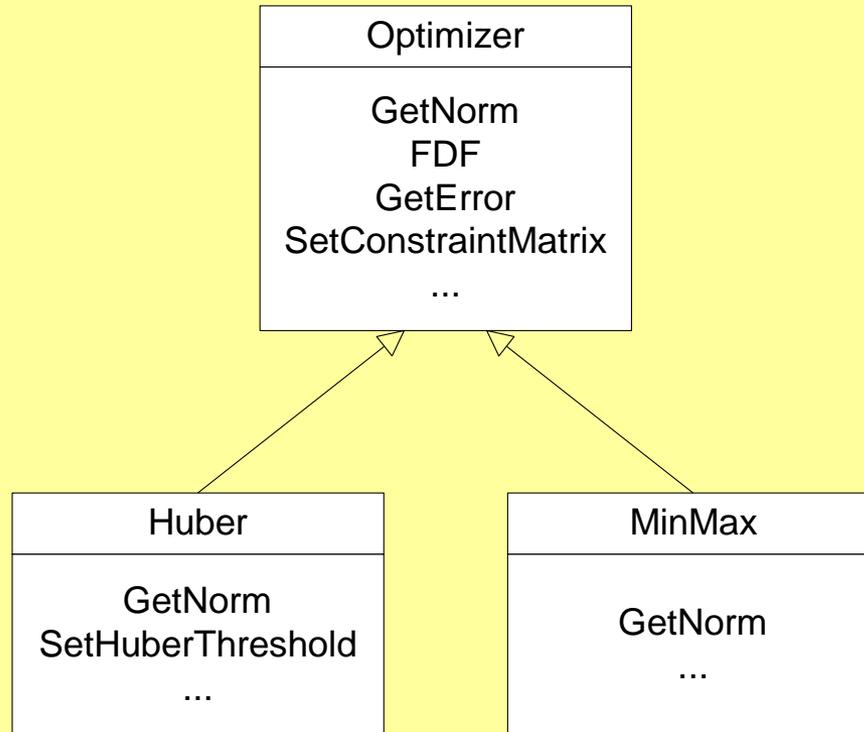
Object Oriented SMX System Design: Data Flow Between Modules





Object Oriented SMX System Design (continued)

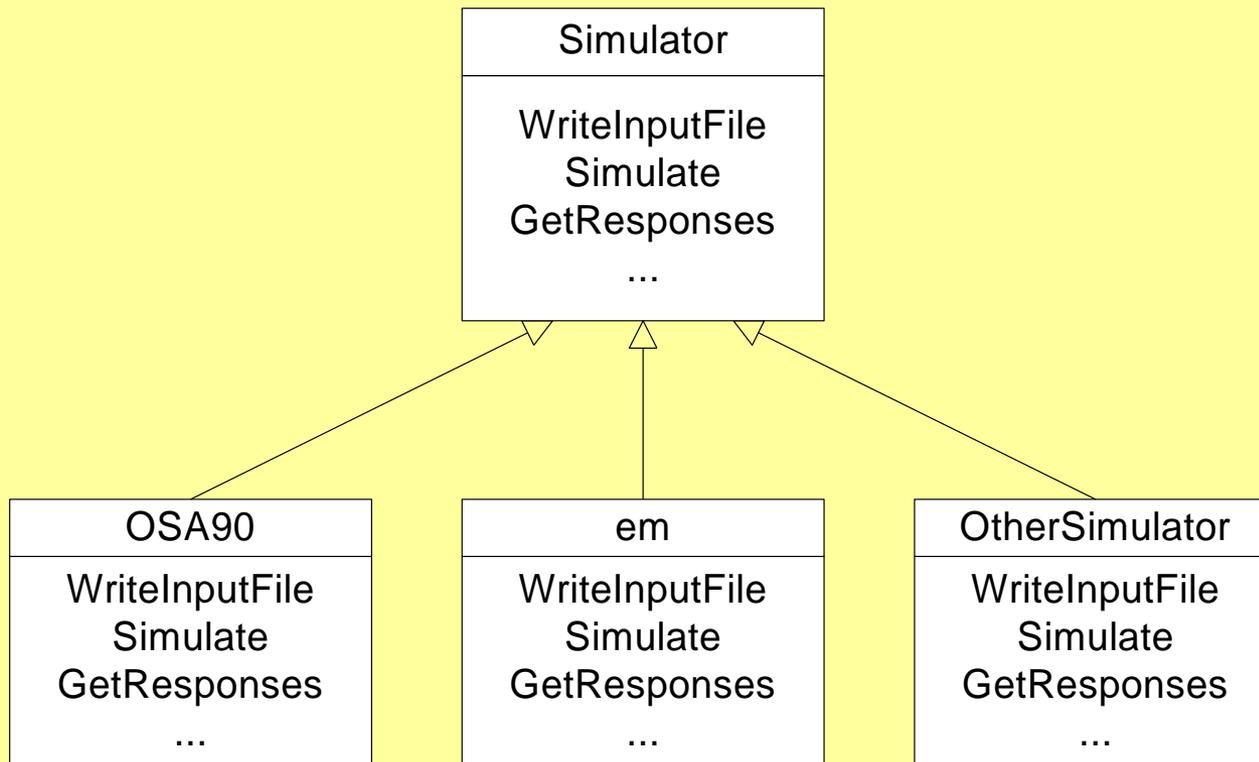
optimizer object: general optimizers





Object Oriented SMX System Design (continued)

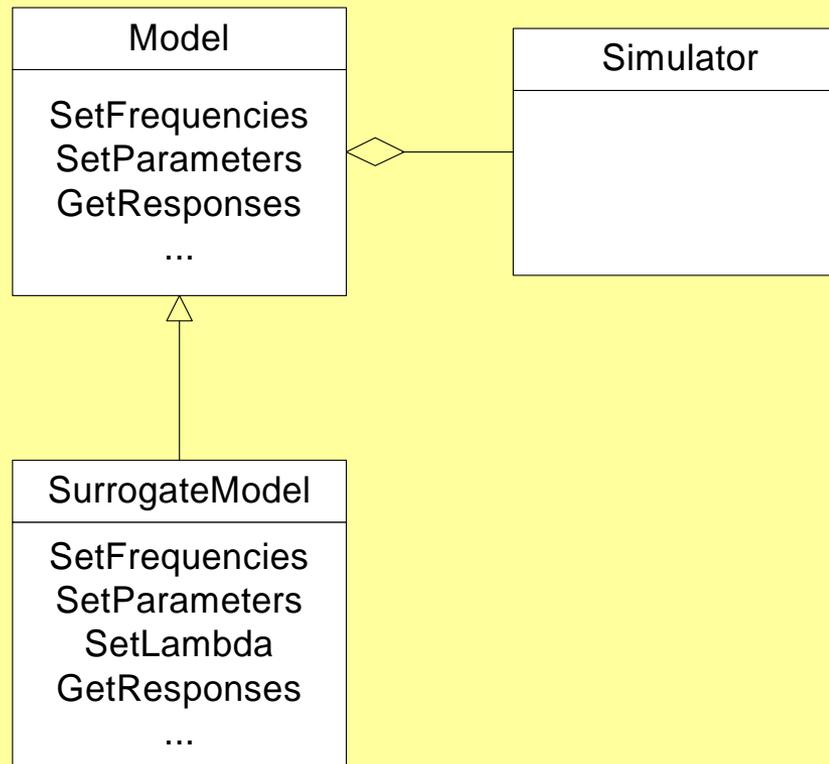
simulator object: interface to simulators





Object Oriented SMX System Design (continued)

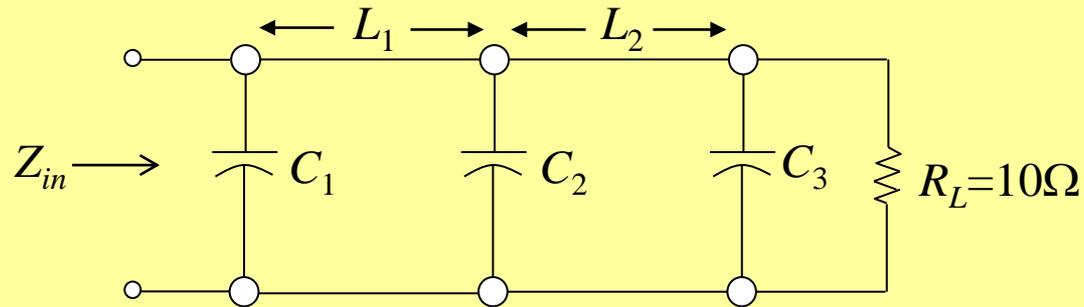
model object: enhanced wrapper of simulators



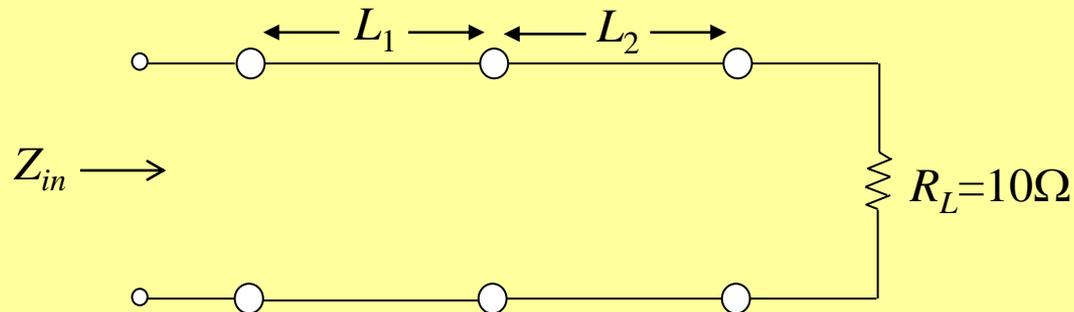


SMX Example: A Two-section 10:1 Capacitively-loaded Impedance Transformer Design

fine model



coarse model



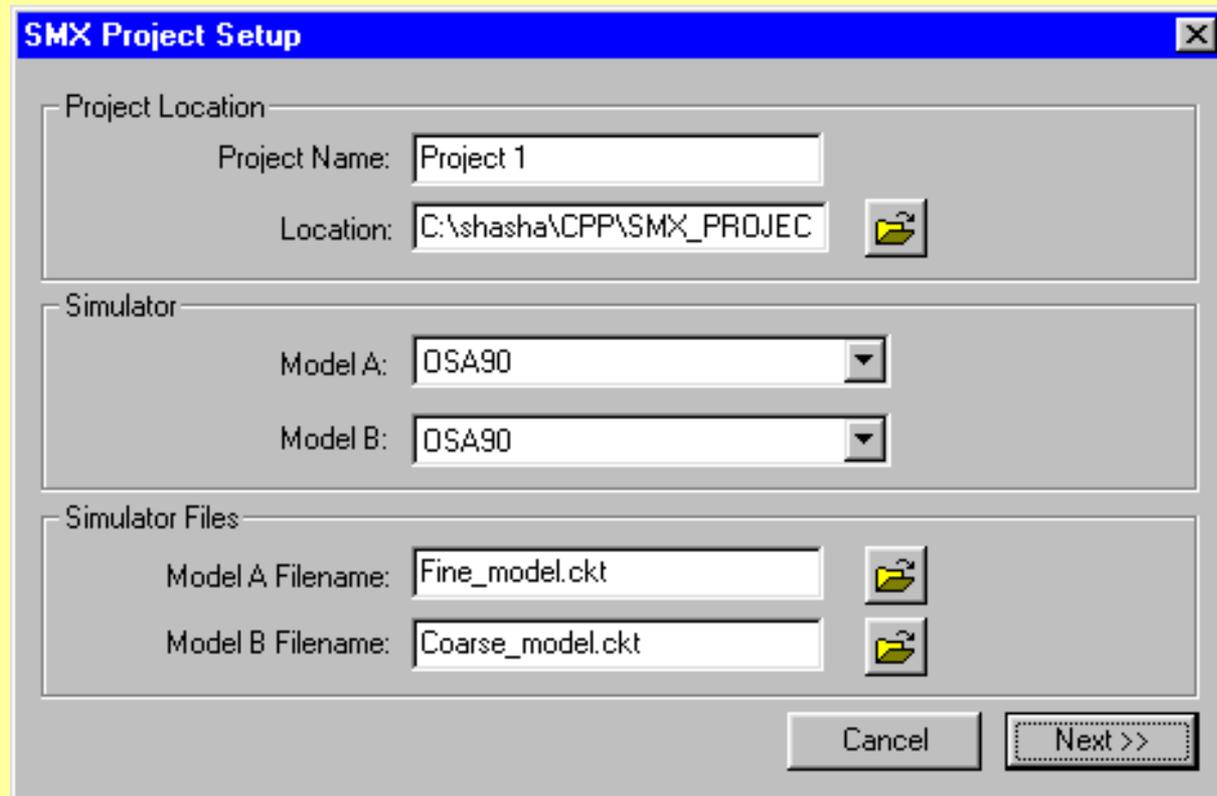
specifications

$$|S_{11}| \leq 0.50 \text{ for } 0.5 \text{ GHz} \leq \omega \leq 1.5 \text{ GHz}$$



Object Oriented SMX System Example: Problem Setup Wizard

step 1: project setup



The image shows a Windows-style dialog box titled "SMX Project Setup". It is divided into three sections: "Project Location", "Simulator", and "Simulator Files".

- Project Location:** Contains a "Project Name" field with the text "Project 1" and a "Location" field with the text "C:\shasha\CPP\SMX_PROJEC". To the right of the location field is a folder icon.
- Simulator:** Contains two dropdown menus. "Model A:" is set to "OSA90" and "Model B:" is also set to "OSA90".
- Simulator Files:** Contains two text fields. "Model A Filename:" is "Fine_model.ckt" and "Model B Filename:" is "Coarse_model.ckt". Each field has a folder icon to its right.

At the bottom right of the dialog box are two buttons: "Cancel" and "Next >>".



Object Oriented SMX System Example: Problem Setup Wizard

step 2: responses and specifications

SMX Responses & Specifications

Specification Setup

Start: 0.5 Stop: 1.5 Step: 0.1 Response: ms11 Constraint: <= Value: 0.5 Add

Specifications

Start	Stop	Step	Response	Constraint	Value
0.500000	1.500000	0.100000	ms11	<=	0.500000

Delete

Mapping Parameters

- Designable Parameters
- Space
- Space and Frequency (decoupled)
- Space and Frequency (coupled)

Simulation Setup

Number of Iterations: 10

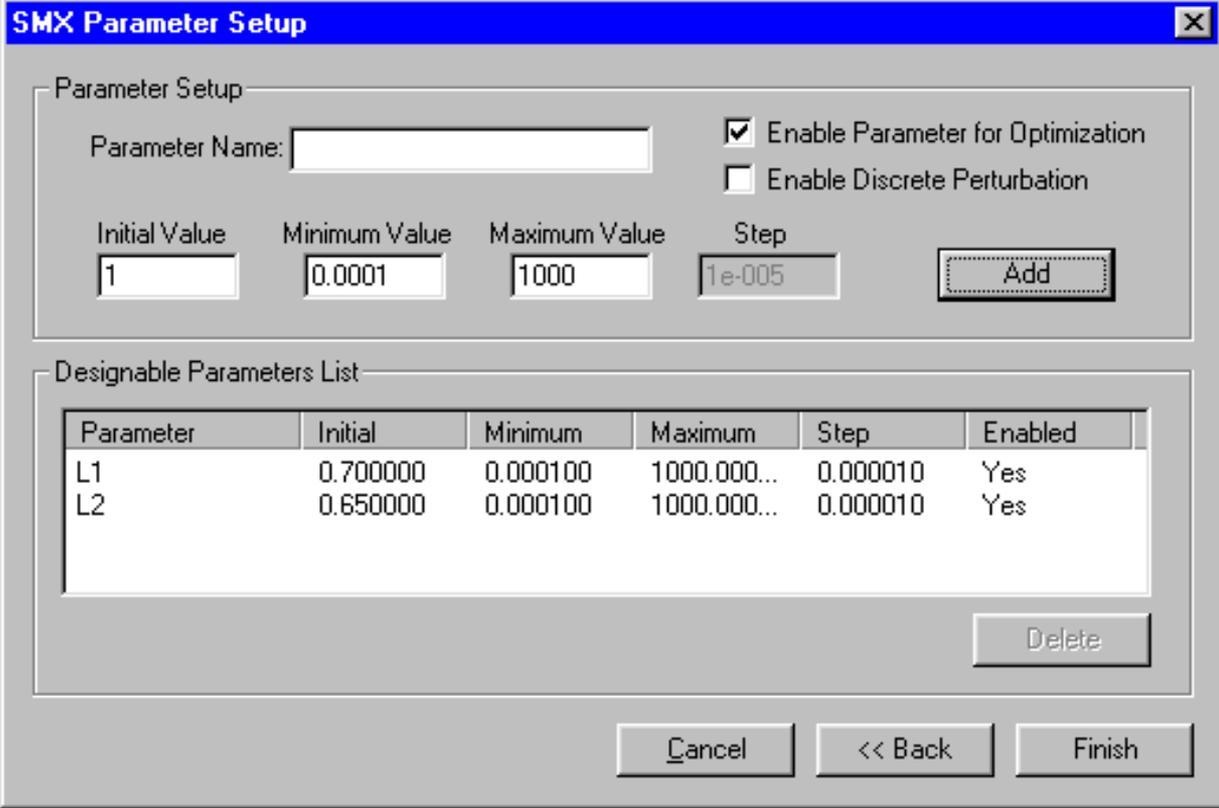
Number of Ports: 1

Cancel << Back Next >>



Object Oriented SMX System Example: Problem Setup Wizard

step 3: parameter setup



The dialog box is titled "SMX Parameter Setup" and contains two main sections: "Parameter Setup" and "Designable Parameters List".

Parameter Setup:

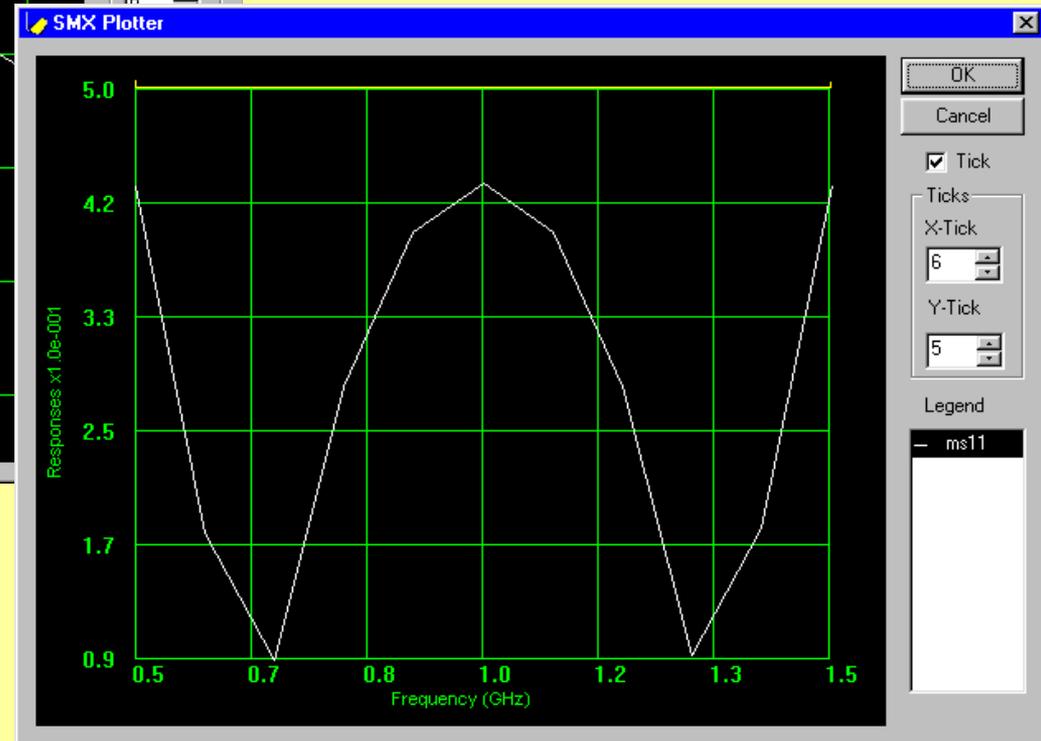
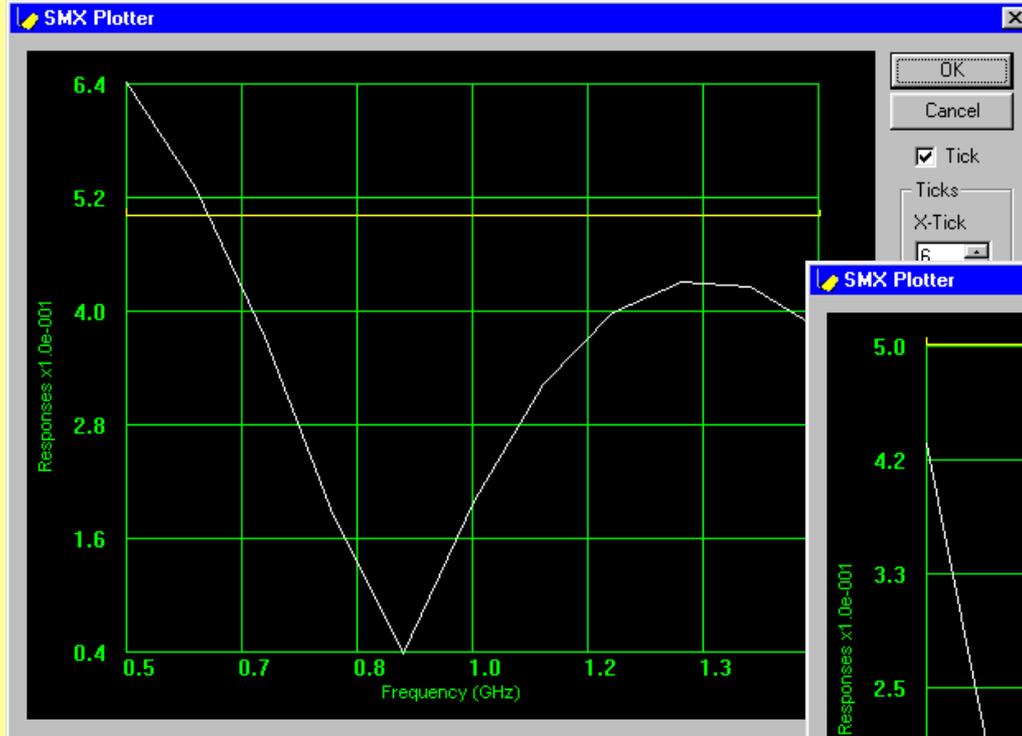
- Parameter Name:
- Enable Parameter for Optimization
- Enable Discrete Perturbation
- Initial Value:
- Minimum Value:
- Maximum Value:
- Step:
-

Designable Parameters List:

Parameter	Initial	Minimum	Maximum	Step	Enabled
L1	0.700000	0.000100	1000.000...	0.000010	Yes
L2	0.650000	0.000100	1000.000...	0.000010	Yes

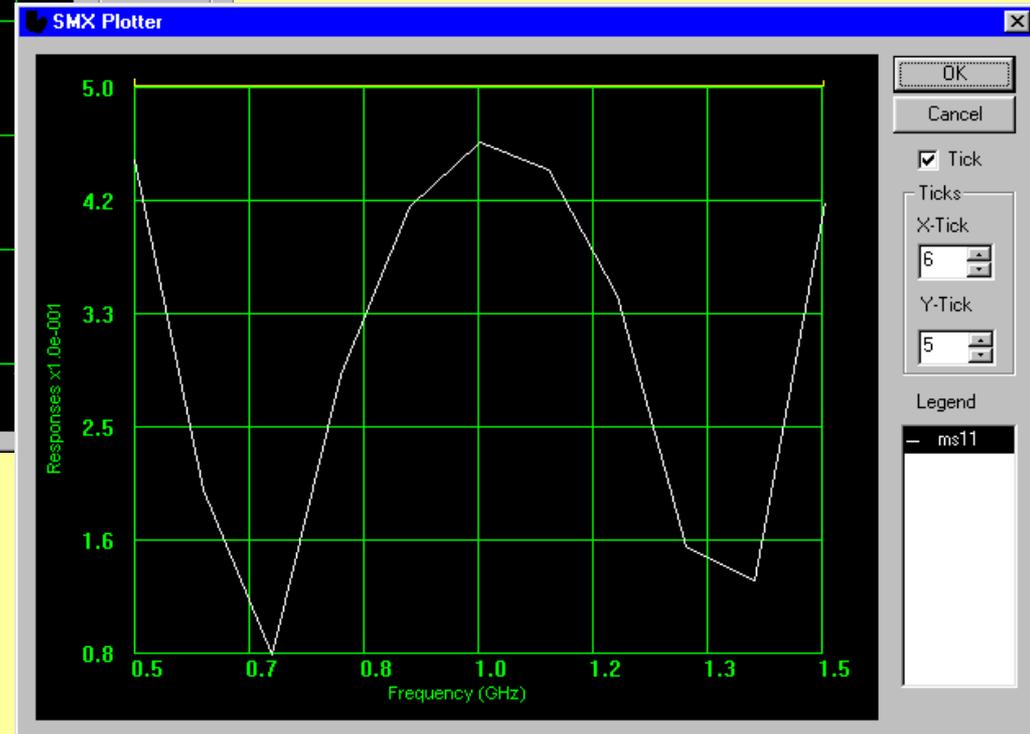
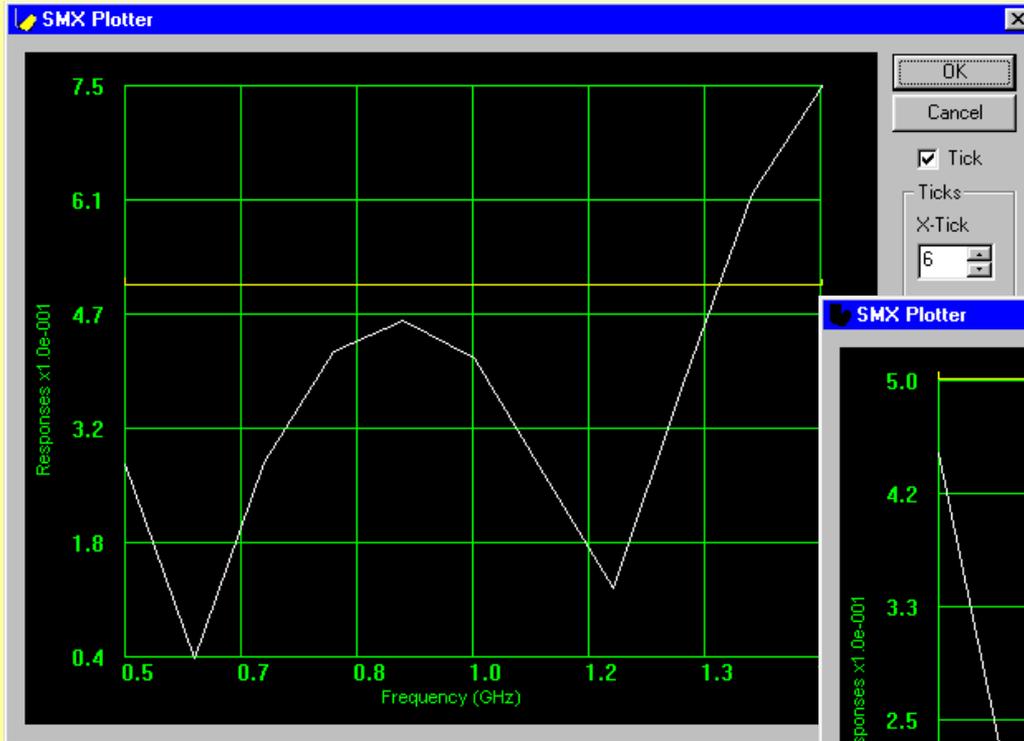


Object Oriented SMX System Example: Initial and Optimal Coarse Model Responses



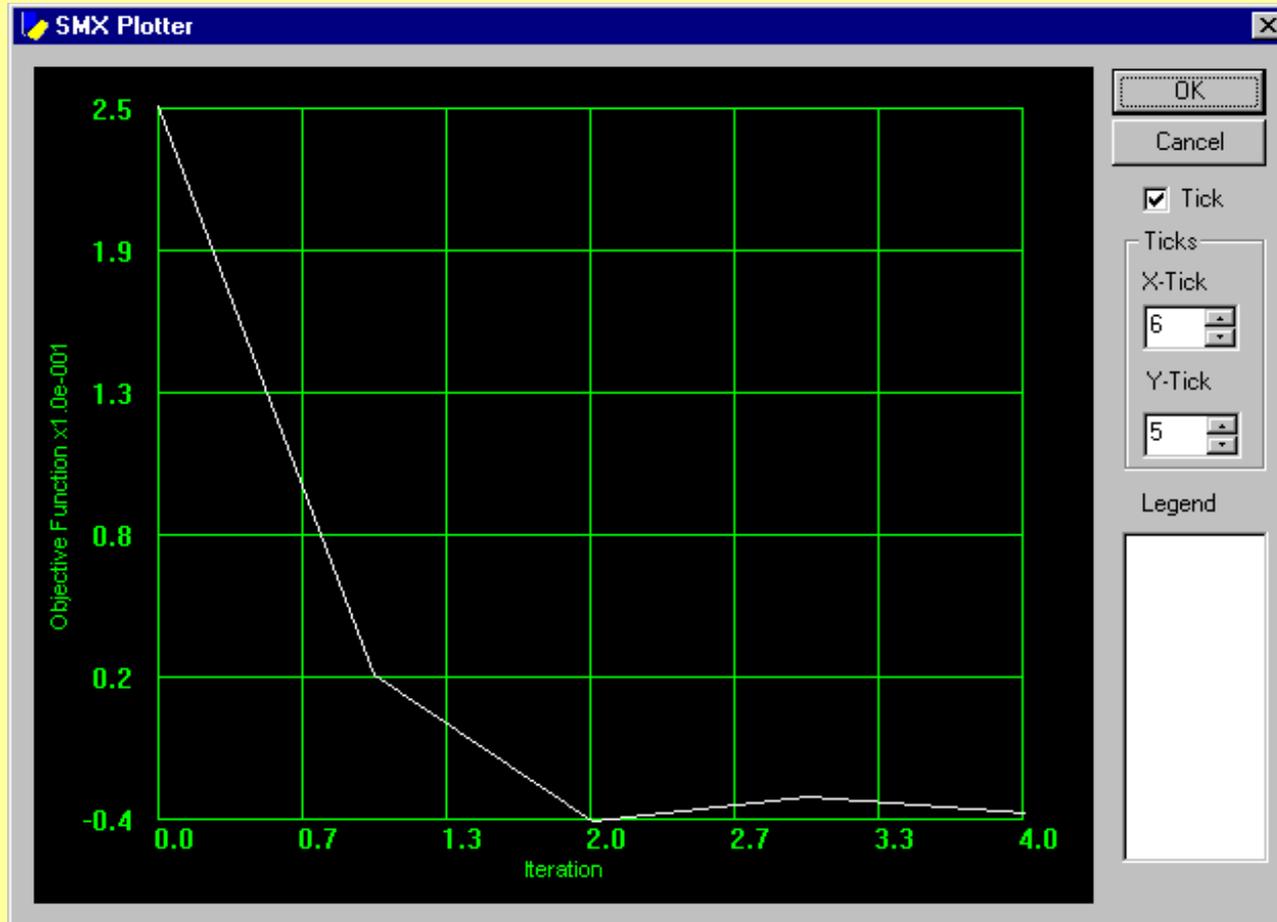


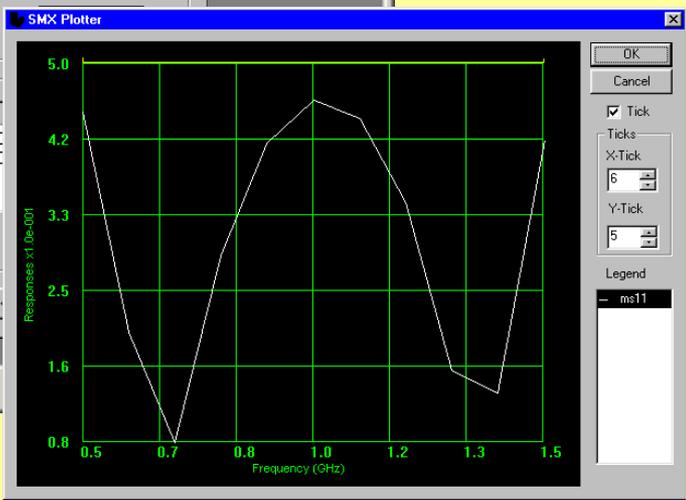
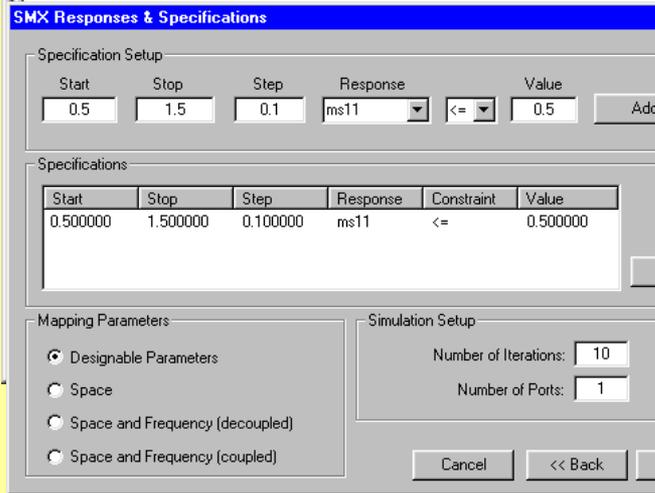
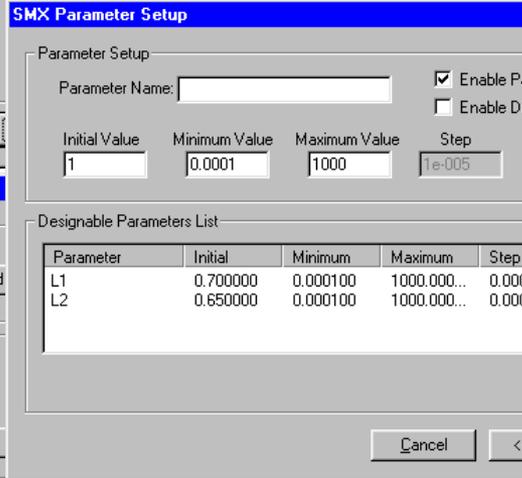
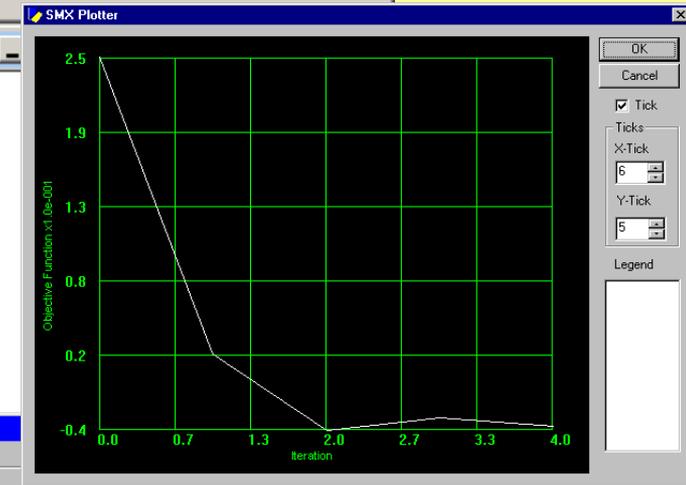
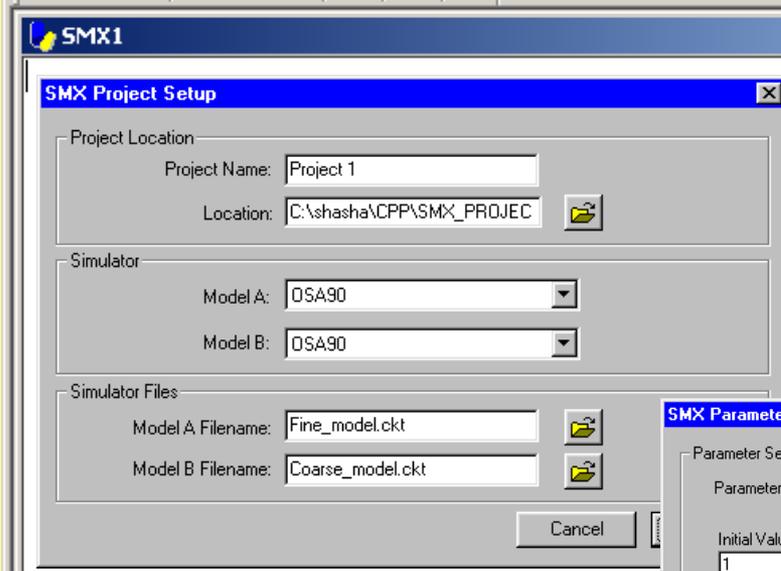
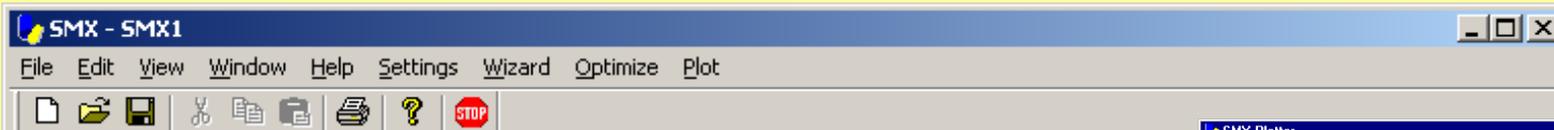
Object Oriented SMX System Example: Initial and Optimal Fine Model Responses





Object Oriented SMX System Example: Objective Function Value







Conclusions

we review Generalized Space Mapping (GSM) as a new engineering device modeling framework that exploits Frequency Space Mapping (FSM) and Multiple Space Mapping (MSM)

we review an innovative algorithm for EM optimization based on Space Mapping technology and Artificial Neural Networks

Neural Space Mapping (NSM) optimization exploits our SM-based neuromodeling techniques

a novel SM optimization algorithm based on surrogate models is presented

the surrogate model is a convex combination of a mapped coarse model and a linearized fine model

the state-of-the-art SMX engineering optimization system including Space Mapping technology is briefly reviewed



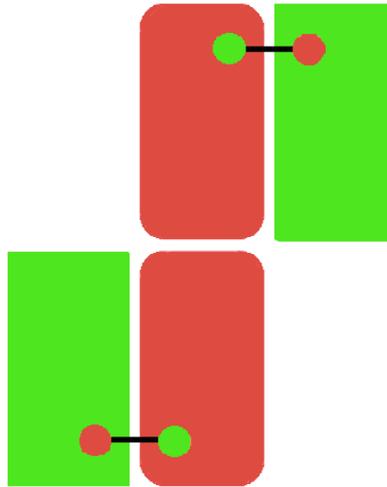
References

- J.W. Bandler, R.M. Biernacki, S.H. Chen and Q.H. Wang, "Multiple space mapping EM optimization of signal integrity in high-speed digital circuits," *Proc. 5th Int. Workshop on Integrated Nonlinear Microwave and Millimeterwave Circuits* (Duisburg, Germany), 1998, pp. 138-140.
- J.W. Bandler, N. Georgieva, M.A. Ismail, J.E. Rayas-Sánchez and Q.J. Zhang, "A generalized space mapping tableau approach to device modeling," *29th European Microwave Conf.* (Munich, Germany), vol. 3, 1999, pp. 231-234.
- J.W. Bandler and J.E. Rayas-Sánchez, "Circuit CAD and modeling through space mapping," *IEEE MTT-S Int. Microwave Symp.*, Workshop WSFD (Anaheim, CA), 1999.
- J.W. Bandler, R.M. Biernacki, S.H. Chen, R.H. Hemmers and K. Madsen, "Electromagnetic optimization exploiting aggressive space mapping," *IEEE Trans. Microwave Theory Tech.*, vol. 43, 1995, pp. 2874-2882.
- M.H. Bakr, J.W. Bandler, R.M. Biernacki, S.H. Chen and K. Madsen, "A trust region aggressive space mapping algorithm for EM optimization," *IEEE Trans. Microwave Theory Tech.*, vol. 46, 1998, pp. 2412-2425.
- M.H. Bakr, J.W. Bandler, N. Georgieva and K. Madsen, "A hybrid aggressive space mapping algorithm for EM optimization," *IEEE Trans. Microwave Theory Tech.*, vol. 47, 1999, pp. 2440-2449.
- A.J. Booker, J.E. Dennis, Jr., P.D. Frank, D. B. Serafini, V. Torczon and M.W. Trosset, "A rigorous framework for optimization of expensive functions by surrogates," *Structural Optimization*, vol. 17, 1999, pp. 1-13.
- V. Torczon and M.W. Trosset, "Using approximations to accelerate engineering design optimization," *Technical Report 98-33, ICASE*, Langley Research Center, Hampton, Virginia 23681-2199, 1998.
- N. Alexandrov, J.E. Dennis, Jr., R.M. Lewis and V. Torczon, "A trust region framework for managing the use of approximation models in optimization," *Structural Optimization*, vol. 15, 1998, pp. 16-23.



References (continued)

- J.W. Bandler, M.A. Ismail, J.E. Rayas-Sánchez and Q.J. Zhang, “Neuromodeling of microwave circuits exploiting space mapping technology,” *IEEE Trans. Microwave Theory Tech.*, vol. 47, 1999, pp. 2417-2427.
- J.W. Bandler, R.M. Biernacki, S.H. Chen, P.A. Grobelny and R.H. Hemmers, “Space mapping technique for electromagnetic optimization,” *IEEE Trans. Microwave Theory Tech.*, vol. 42, 1994, pp. 2536-2544.
- J.W. Bandler and S.H. Chen, “Circuit optimization: the state of the art,” *IEEE Trans. Microwave Theory Tech.*, vol. 36, 1988, pp. 424-443.
- J.W. Bandler, R.M. Biernacki, S.H. Chen, W.J. Getsinger, P.A. Grobelny, C. Moskowitz and S.H. Talisa, “Electromagnetic design of high-temperature superconducting microwave filters,” *Int. J. Microwave and Millimeter-Wave CAE*, vol. 5, 1995, pp. 331-343.
- em*TM Version 5.1a, Sonnet Software, Inc., 1020 Seventh North Street, Suite 210, Liverpool, NY 13088, 1998.
- EmpipeTM Version 4.0, formerly Optimization Systems Associates Inc., P.O. Box 8083, Dundas, Ontario, Canada L9H 5E7, 1997, now Agilent EEsof EDA, 1400 Fountaingrove Parkway, Santa Rosa, CA 95403-1799.
- OSA90/hopeTM Version 4.0, formerly Optimization Systems Associates Inc., P.O. Box 8083, Dundas, Ontario, Canada L9H 5E7, 1997, now Agilent EEsof EDA, 1400 Fountaingrove Parkway, Santa Rosa, CA 95403-1799.
- J.W. Bandler, M.A. Ismail and J.E. Rayas-Sánchez, “Broadband physics-based modeling of microwave passive devices through frequency mapping,” *IEEE MTT-S Int. Microwave Symp. Digest* (Boston, MA), 2000, pp. 969-972.
- M.H. Bakr, J.W. Bandler, K. Madsen, J.E. Rayas-Sánchez and J. Søndergaard, “Space mapping optimization of microwave circuits exploiting surrogate models,” *IEEE MTT-S Int. Microwave Symp. Digest* (Boston, MA), 2000, pp. 1785-1788.
- M.H. Bakr, J.W. Bandler, M.A. Ismail, J.E. Rayas-Sánchez and Q.J. Zhang, “Neural space mapping optimization of EM microwave structures,” *IEEE MTT-S Int. Microwave Symp. Digest* (Boston, MA), 2000, pp. 879-882.



First International Workshop on
SURROGATE MODELLING AND SPACE MAPPING
FOR ENGINEERING OPTIMIZATION

Technical University of Denmark
Lyngby, Denmark
November 16-18, 2000