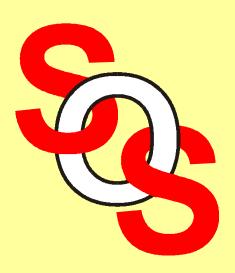
SPACE MAPPING TECHNOLOGY: A NEW APPROACH TO ENGINEERING OPTIMIZATION

John W. Bandler

Simulation Optimization Systems Research Laboratory and Department of Electrical and Computer Engineering McMaster University, Hamilton, Canada L8S 4K1

> Bandler Corporation, john@bandler.com http://www.bandler.com



presented at

SPACE MAPPING TECHNOLOGY: A NEW APPROACH TO ENGINEERING OPTIMIZATION

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Abstract

Electromagnetics (EM) based device modeling and circuit optimization through Space Mapping (SM) technologies are reviewed. The SM concept continues to promise important benefits in the next generation of design optimization methodologies. Artificial Neural Networks can be incorporated into the SM optimization strategies. Aggressive Space Mapping (ASM) optimization closely follows the traditional experience and intuition of designers, while being rigorously grounded mathematically. Current progress in the development of suitable algorithms and software engines are presented. The SM concept addresses the contradictory challenge of exploitation of device models for CAD that are both accurate and fast.

Outline

a comprehensive Generalized Space Mapping (GSM) tableau approach (Bandler et al., 1999) to engineering device modeling exploiting Frequency Space Mapping (FSM) (Bandler et al., 1995) and the Multiple Space Mapping (MSM) (Bandler et al., 1998) is reviewed

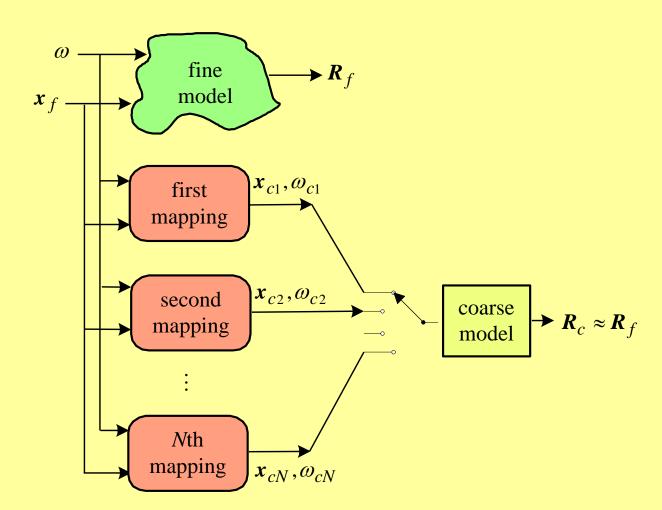
a Neural Space Mapping (NSM) optimization approach exploiting our SM-based neuromodeling techniques is presented (*Bakr et al.*, 2000)

new work on Space Mapping optimization exploiting surrogate models is described (Bakr et al., 2000)

a state-of-the-art engineering optimization system including the latest Space Mapping technology, the SMX system, is described (*Bandler et al.*, 2000)

Multiple Space Mapping (MSM) Concept

MSM for Frequency Intervals (MSMFI)



Mathematical Formulation for GSM

(*Bandler et al.*, 1999)

the kth mapping targeting the sub-response or the response R in the kth frequency sub-range is given by

$$(\boldsymbol{x}_{ck}, \omega_{ck}) = \boldsymbol{P}_k(\boldsymbol{x}_f, \omega)$$

or, in matrix form, assuming a linear mapping

$$\begin{bmatrix} \boldsymbol{x}_{ck} \\ \omega_{ck} \end{bmatrix} = \begin{bmatrix} \boldsymbol{c}_k \\ \delta_k \end{bmatrix} + \begin{bmatrix} \boldsymbol{B}_k & \boldsymbol{s}_k \\ \boldsymbol{t}_k^T & \sigma_k \end{bmatrix} \begin{bmatrix} \boldsymbol{x}_f \\ \omega \end{bmatrix}$$

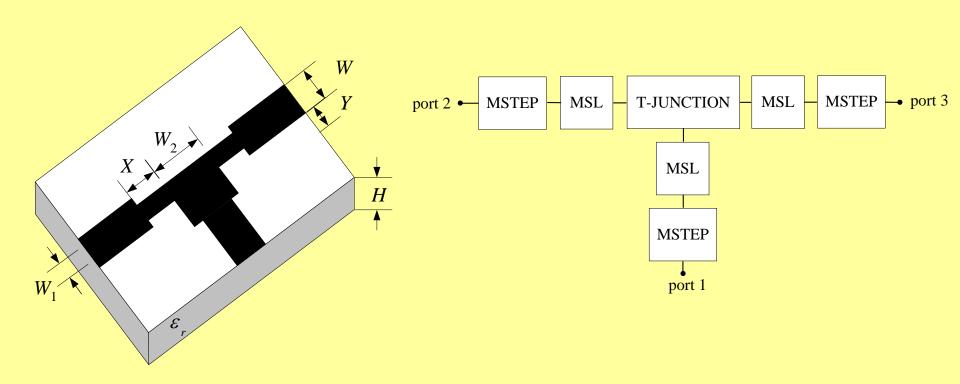
the mapping parameters $\{c_k, B_k, s_k, t_k, \sigma_k, \delta_k\}$ can be evaluated, directly or indirectly, by solving the optimization problem

$$\min_{\boldsymbol{c}_k,\,\boldsymbol{B}_k,\,\boldsymbol{s}_k,\,\boldsymbol{t}_k,\,\sigma_k,\,\delta_k} \left\| [\boldsymbol{e}_{k1}^T \quad \boldsymbol{e}_{k2}^T \quad \cdots \quad \boldsymbol{e}_{km}^T]^T \right\|$$

where m is the number of base points selected in the fine model space and e_{kj} is an error vector given by

$$\boldsymbol{e}_{kj} = \boldsymbol{R}_f(\boldsymbol{x}_f^{(j)}, \omega) - \boldsymbol{R}_c(\boldsymbol{x}_{ck}^{(j)}, \omega_{ck}), \quad j = 1, 2, ..., m$$

the fine and coarse models



the region of interest

15 mil
$$\leq H \leq$$
 25 mil
5 mil $\leq X \leq$ 15 mil
5 mil $\leq Y \leq$ 15 mil
 $8 \leq \varepsilon_r \leq$ 10

the frequency range is 2 GHz to 20 GHz with a step of 2 GHz

the number of base points is 9 and the number of test points is 50

the width W of the input lines is determined in terms of H and so that the characteristic impedance of the input lines is 50 ohm

the width W_1 is taken as 1/3 of the width W

the width W_2 is obtained so that the characteristic impedance of the microstrip line after the step connected to port 2 is twice that of the microstrip line after the step connected to port 1

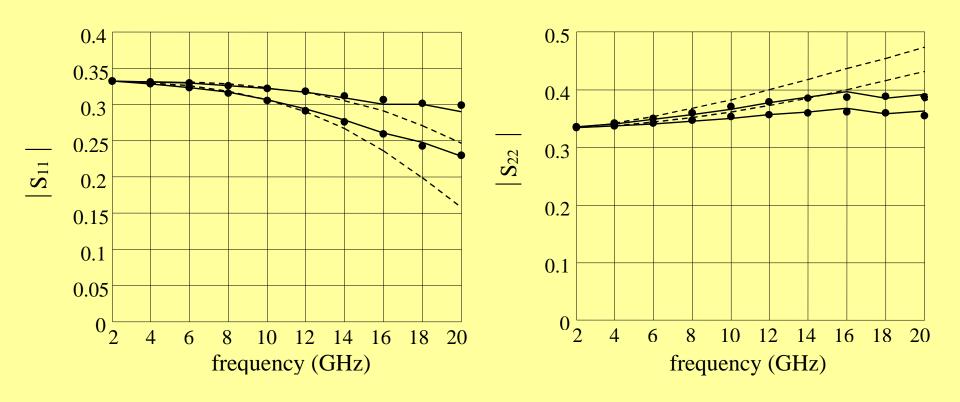
MSM for Frequency Intervals (MSMFI) was developed to enhance the accuracy of the T-Junction coarse model

the total frequency range was divided into two intervals: 2-16 GHz and 16-20 GHz

the mapping parameters are

	2 GHz to 16 GHz	16 GHz to 20 GHz
В	$\begin{bmatrix} 1.04 & 0.07 & 0.01 & 0.08 - 0.06 & 0.00 & 0.22 \\ 0.00 & 0.89 & 0.00 - 0.07 - 0.20 & 0.06 - 0.03 \\ -0.00 & 0.07 & 0.99 & 0.04 - 0.12 & 0.01 - 0.06 \\ -0.04 & 0.00 - 0.01 & 0.97 & 0.10 - 0.06 - 0.27 \\ 0.01 & 0.04 & 0.00 & 0.03 & 0.99 - 0.05 - 0.03 \\ -0.13 - 0.05 & -0.04 - 0.16 & 0.12 & 0.99 & 0.62 \\ -0.08 & 0.12 & -0.03 & 0.00 & -0.07 & 0.03 & 0.83 \end{bmatrix}$	$\begin{bmatrix} 0.99 & 0.02 - 0.00 & 0.01 - 0.09 - 0.01 & 0.13 \\ 0.05 & 0.85 & 0.01 - 0.07 - 0.28 & 0.01 - 0.01 \\ -0.06 & 0.15 & 0.98 & 0.04 - 0.25 & 0.00 & 0.02 \\ -0.10 - 0.06 - 0.03 & 0.88 & 0.13 - 0.09 - 0.27 \\ 0.08 & 0.04 & 0.03 & 0.11 & 1.07 - 0.04 - 0.12 \\ -0.14 - 0.02 - 0.05 - 0.15 & 0.23 & 1.03 & 0.51 \\ -0.13 & 0.22 - 0.04 & 0.02 - 0.07 & 0.03 & 0.87 \end{bmatrix}$
c	$[0.02 0.01 -0.01 -0.03 -0.01 0.07 -0.03]^T$	$\begin{bmatrix} 0.01 & 0.01 & -0.01 & -0.03 & -0.01 & 0.05 & -0.03 \end{bmatrix}^T$
S	$\begin{bmatrix} -0.01 & 0.09 & -0.10 & -0.02 & 0.00 & -0.02 & -0.20 \end{bmatrix}^T$	$\begin{bmatrix} 0.00 & 0.01 & -0.01 & 0.00 & 0.00 & 0.00 & -0.02 \end{bmatrix}^T$
t	0	$\begin{bmatrix} 0.01 & 0.00 & -0.02 & 0.00 & 0.00 & 0.00 & 0.00 \end{bmatrix}^T$
σ	0.851	0.957
δ	-0.003	0.008

the responses of the shaped T-Junction at two test points in the region of interest by Sonnet's *em* (•), by the coarse model (---) and by the enhanced coarse model (—)



the enhanced coarse model for the shaped T-Junction can be utilized in optimization

the optimization variables are *X* and *Y*

the other parameters are kept fixed (W = 24 mil, H = 25 mil and $\varepsilon_r = 9.9$)

the design specifications are

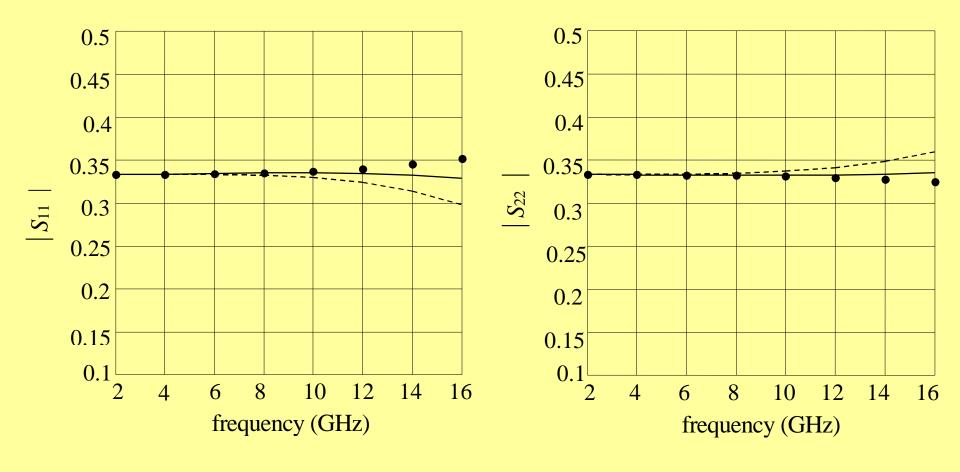
$$|S_{11}| \le 1/3, |S_{22}| \le 1/3$$

in the frequency range 2 GHz to 16 GHz

the minimax optimizer in OSA90/hope reached the solution

$$X = 2.1 \text{ mil and } Y = 21.1 \text{ mil}$$

responses of the optimal shaped T-Junction by Sonnet's *em* (•), by the coarse model (---) and by the enhanced coarse model (—)



Neural Space Mapping (NSM) Optimization

(*Bakr et al.*, 2000)

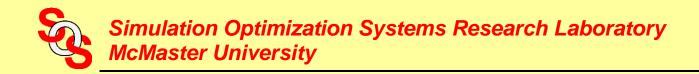
exploits the SM-based neuromodeling techniques (Bandler et al., 1999)

coarse models are used as sources of knowledge that reduce the amount of learning data and improve the generalization and extrapolation performance

NSM requires a reduced set of upfront learning base points

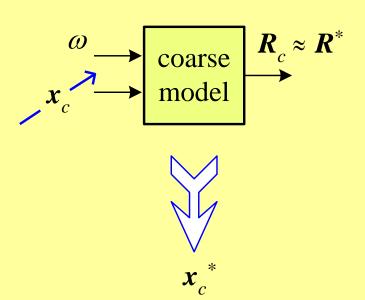
the initial learning base points are selected through sensitivity analysis using the coarse model

neuromappings are developed iteratively: their generalization performance is controlled by gradually increasing their complexity starting with a 3-layer perceptron with 0 hidden neurons

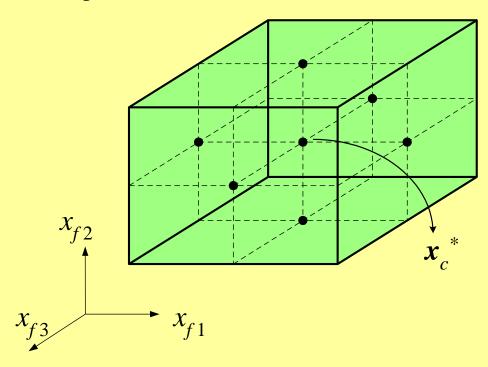


Neural Space Mapping (NSM) Optimization Concept

step 1



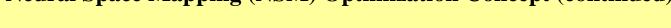
step 2

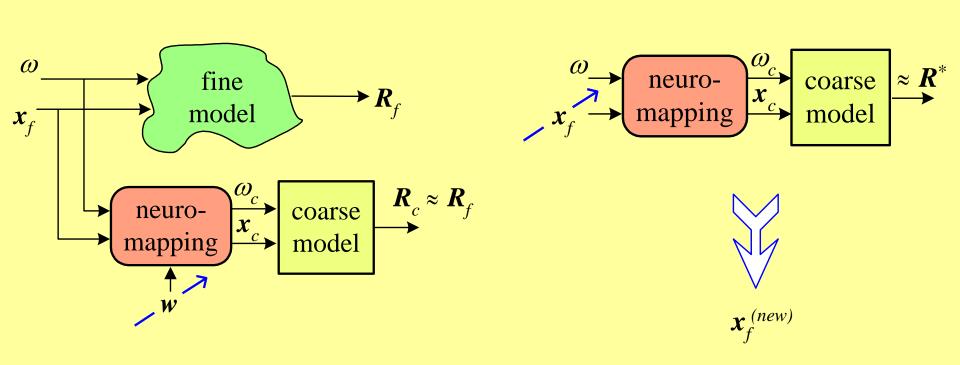


(2n + 1 learning base points for a microwave circuit with n design parameters)

step 3

Neural Space Mapping (NSM) Optimization Concept (continued)





step 4

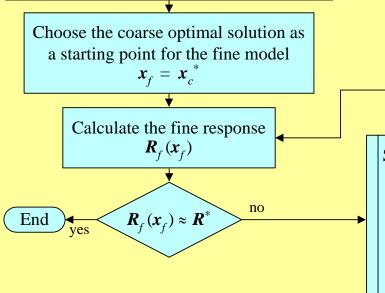
Neural Space Mapping (NSM) Optimization Algorithm



COARSE OPTIMIZATION: find the optimal coarse model solution \mathbf{x}_c^* that generates the desired response \mathbf{R}^*

$$\mathbf{R}_{c}(\mathbf{x}_{c}^{*}) = \mathbf{R}^{*}$$

Form a learning set with $B_p = 2n+1$ base points, by selecting 2n additional points around x_c^* , following a star distribution



Include the new x_f in the learning set and increase B_p by one

SM BASED NEUROMODELING: Find the simplest neuromapping *P* such that

$$\mathbf{R}_{f}(\mathbf{x}_{f}^{(l)}, \omega_{j}) \approx \mathbf{R}_{c}(\mathbf{P}(\mathbf{x}_{f}^{(l)}, \omega_{j}))$$

$$l = 1,..., B_p$$
 and $j = 1,..., F_p$

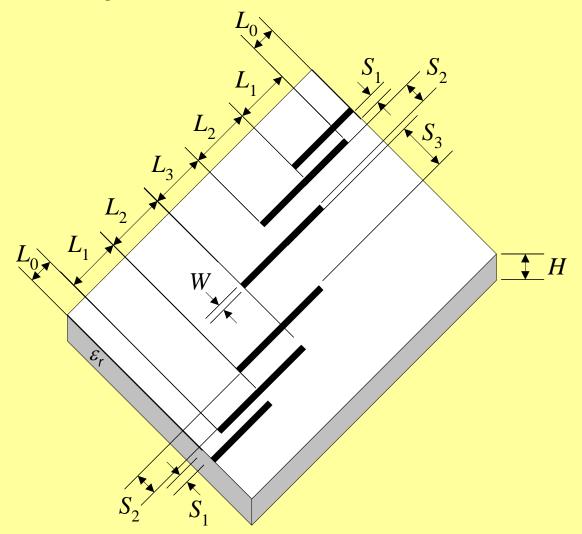
SMBNM OPTIMIZATION: Find the optimal x_f such that

Update x_f

$$\boldsymbol{R}_{SMBN}(\boldsymbol{x}_f) = \boldsymbol{R}_c(\boldsymbol{P}(\boldsymbol{x}_f)) \approx \boldsymbol{R}^*$$

HTS Quarter-Wave Parallel Coupled-Line Microstrip Filter

(Westinghouse, 1993)



we take $L_0 = 50$ mil, H = 20 mil, W = 7 mil, $\varepsilon_r = 23.425$, loss tangent = 3×10^{-5} ; the metalization is considered lossless

the design parameters are $\mathbf{x}_f = [L_1 \ L_2 \ L_3 \ S_1 \ S_2 \ S_3]^T$

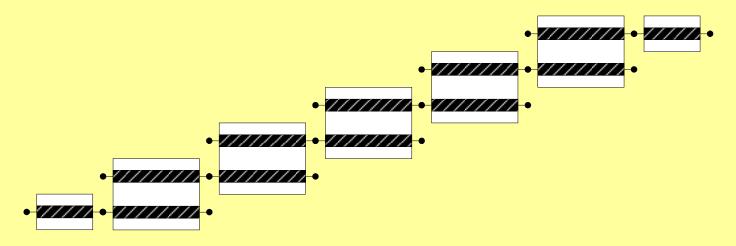
NSM Optimization of the HTS Microstrip Filter

specifications

$$|S_{21}| \ge 0.95$$
 for 4.008 GHz $\le f \le 4.058$ GHz $|S_{21}| \le 0.05$ for $f \le 3.967$ GHz and $f \ge 4.099$ GHz

"fine" model: Sonnet's emTM with high resolution grid

"coarse" model: OSA90/hope™ built-in models of open circuits, microstrip lines and coupled microstrip lines

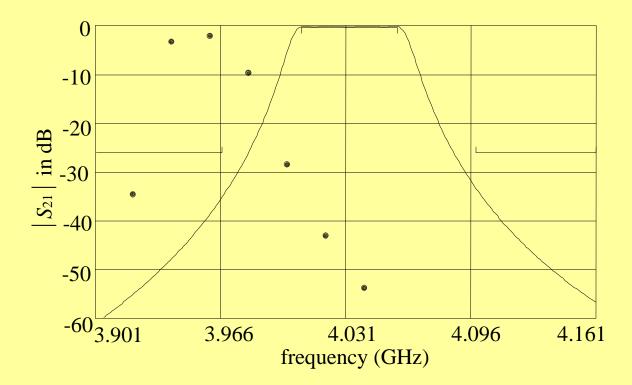


NSM Optimization of the HTS Filter (continued)

coarse and fine model responses at the optimal coarse solution

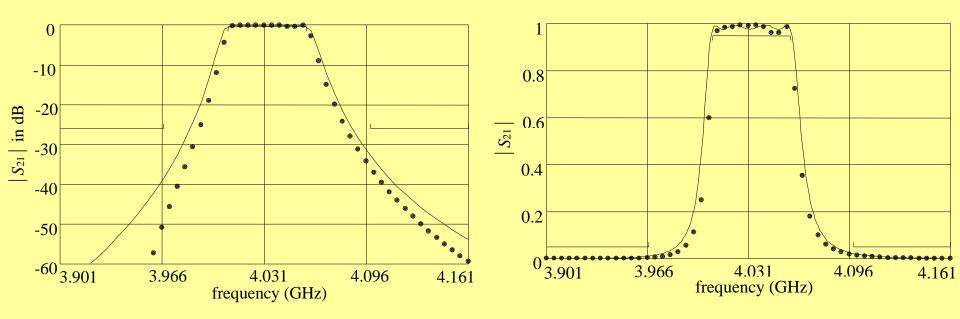
$$\mathbf{x}_{c}^{*} = [188.33 \ 197.98 \ 188.58 \ 21.97 \ 99.12 \ 111.67]^{T}$$
 (mils)

OSA90/hopeTM (-) and em^{TM} (\bullet)



NSM Optimization of the HTS Filter (continued)

em[™] (•) and FPSM 7-5-3 (–) model responses at the NSM solution using a fine frequency sweep



Space Mapping Optimization Exploiting Surrogates

(*Bakr et al.*, 2000)

a powerful new Space Mapping (SM) optimization algorithm has been developed

it draws upon recent developments in both surrogate model-based optimization and modeling of microwave devices

SM optimization is formulated as a general optimization problem of a surrogate model

this model is a convex combination of a mapped coarse model and a linearized fine model

it exploits, in a novel way, a linear frequency-sensitive mapping

during the optimization iterates, the coarse and fine models are simulated at different sets of frequencies.

this approach is shown to be especially powerful if a significant response shift exists

The Surrogate Model

our surrogate model is a convex combination of a mapped coarse model and a linearized fine model

the *i*th iteration surrogate model is

$$\mathbf{R}_{s}^{(i)}(\mathbf{x}_{f}) = \lambda^{(i)} \mathbf{R}_{m}^{(i)}(\mathbf{x}_{f}) + (1 - \lambda^{(i)}) (\mathbf{R}_{f}(\mathbf{x}_{f}^{(i)}) + \mathbf{J}_{f}^{(i)} \Delta \mathbf{x}_{f}), \quad \lambda^{(i)} \in [0, 1]$$

the mapped coarse model utilizes the frequency-sensitive mapping

$$\mathbf{R}_f(\mathbf{x}_f, \omega_j) \approx \mathbf{R}_m^{(i)}(\mathbf{x}_f, \omega_j) = \mathbf{R}_c(\mathbf{P}^{(i)}(\mathbf{x}_f, \omega_j), \mathbf{P}_\omega^{(i)}(\mathbf{x}_f, \omega_j))$$

where

$$\begin{bmatrix} \mathbf{P}^{(i)}(\mathbf{x}_f, \omega_j) \\ \mathbf{P}^{(i)}_{\omega}(\mathbf{x}_f, \omega_j) \end{bmatrix} = \begin{bmatrix} \mathbf{B}^{(i)} & \mathbf{s}^{(i)} \\ \mathbf{t}^{(i)T} & \sigma^{(i)} \end{bmatrix} \begin{bmatrix} \Delta \mathbf{x}_f \\ \omega_j \end{bmatrix} + \begin{bmatrix} \mathbf{c}^{(i)} \\ \mathbf{y}^{(i)} \end{bmatrix}$$

the parameters $\mathbf{B}^{(i)} \in \mathbb{R}^{n \times n}$, $\mathbf{s}^{(i)} \in \mathbb{R}^{n \times 1}$, $\mathbf{t}^{(i)} \in \mathbb{R}^{n \times 1}$, $\mathbf{c}^{(i)} \in \mathbb{R}^{n \times 1}$, $\sigma^{(i)} \in \mathbb{R}^{1 \times 1}$ and $\gamma^{(i)} \in \mathbb{R}^{1 \times 1}$ are obtained such that the mapped coarse model approximates the fine model over a given set of fine model points $V^{(i)}$ and frequencies $\boldsymbol{\omega}$

The Surrogate Model (continued)

the mapping parameters are obtained through the optimization process

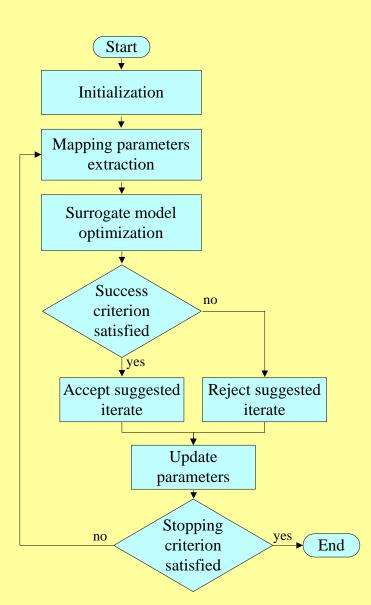
$$[\boldsymbol{B}^{(i)}, \boldsymbol{s}^{(i)}, \boldsymbol{t}^{(i)}, \sigma^{(i)}, \boldsymbol{c}^{(i)}, \gamma^{(i)}] =$$

$$\operatorname{arg} \left\{ \min_{\boldsymbol{B}, \boldsymbol{s}, \boldsymbol{t}, \sigma, \boldsymbol{c}, \gamma} \left\| [\boldsymbol{e}_{1}^{T} \quad \boldsymbol{e}_{2}^{T} \quad \cdots \quad \boldsymbol{e}_{N_{p}}^{T}]^{T} \right\| \right\}$$

where

$$\boldsymbol{e}_{k} = \boldsymbol{R}_{m}^{(i)}(\boldsymbol{x}_{f}^{(k)}) - \boldsymbol{R}_{f}(\boldsymbol{x}_{f}^{(k)}) \qquad \forall \, \boldsymbol{x}_{f}^{(k)} \in \boldsymbol{V}^{(i)}$$

The Algorithm Flowchart



The SMX System

(*Bandler et al., 2000*)

SMX is a new generation engineering optimization system

currently it provides the following optimization capabilities

minimax

Huber

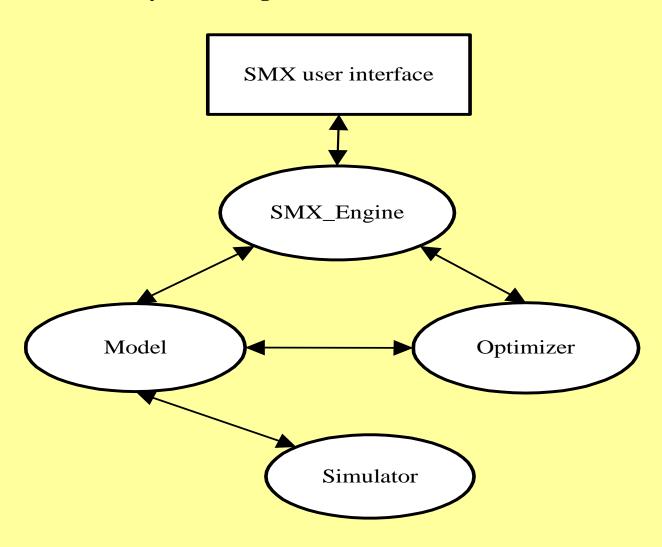
Space Mapping using Surrogate Models (Bakr et al., 2000)

currently it can be interfaced to

OSA90

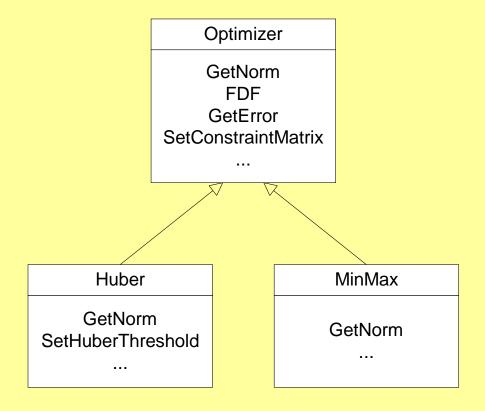
user supplied executable programs

Object Oriented SMX System Design: Data Flow Between Modules



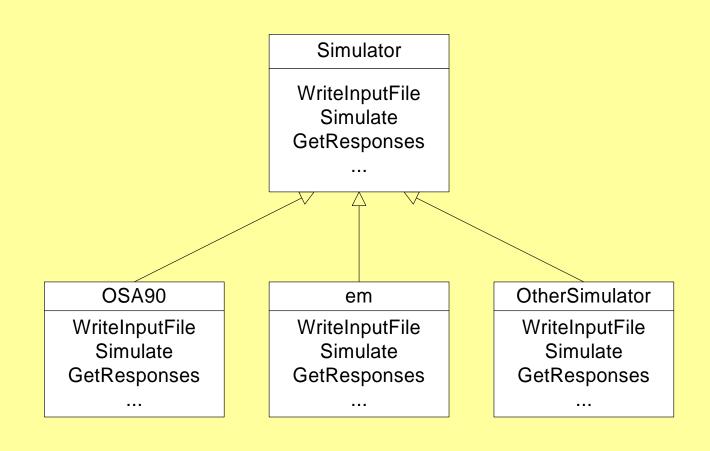
Object Oriented SMX System Design (continued)

optimizer object: general optimizers



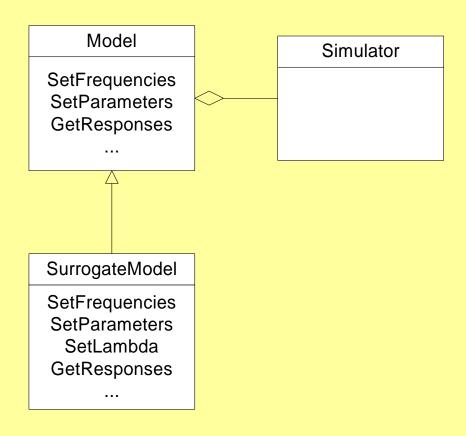
Object Oriented SMX System Design (continued)

simulator object: interface to simulators



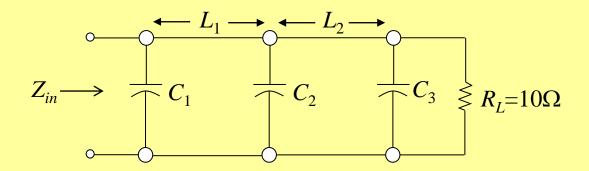
Object Oriented SMX System Design (continued)

model object: enhanced wrapper of simulators

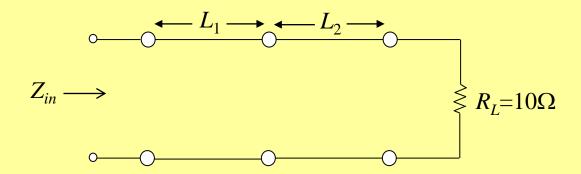


SMX Example: A Two-section 10:1 Capacitively-loaded Impedance Transformer Design

fine model



coarse model

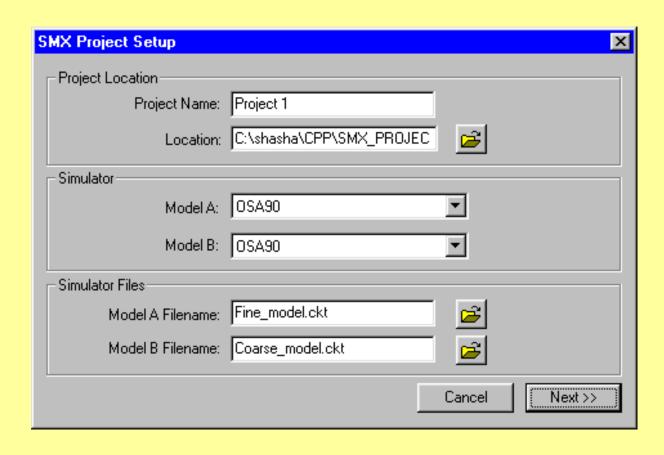


specifications

$$|S_{11}| \le 0.50$$
 for $0.5 \text{ GHz} \le \omega \le 1.5 \text{ GHz}$

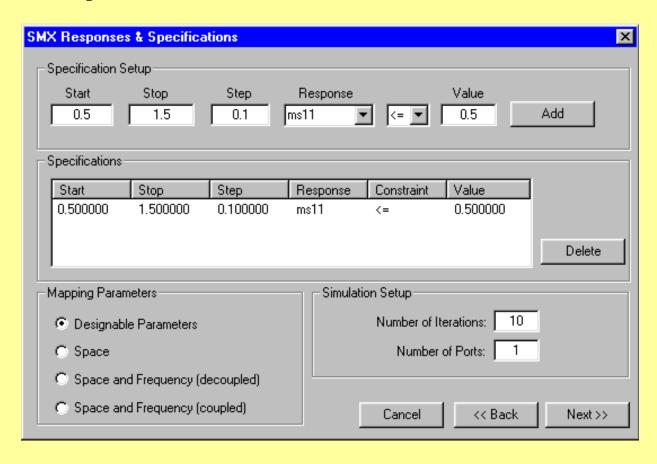
Object Oriented SMX System Example: Problem Setup Wizard

step 1: project setup



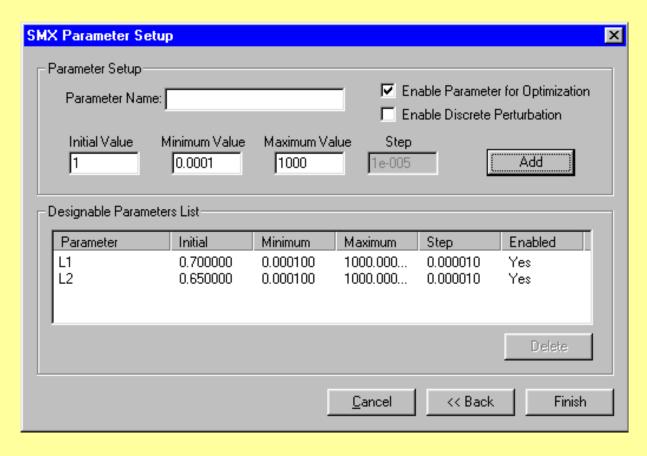
Object Oriented SMX System Example: Problem Setup Wizard

step 2: responses and specifications

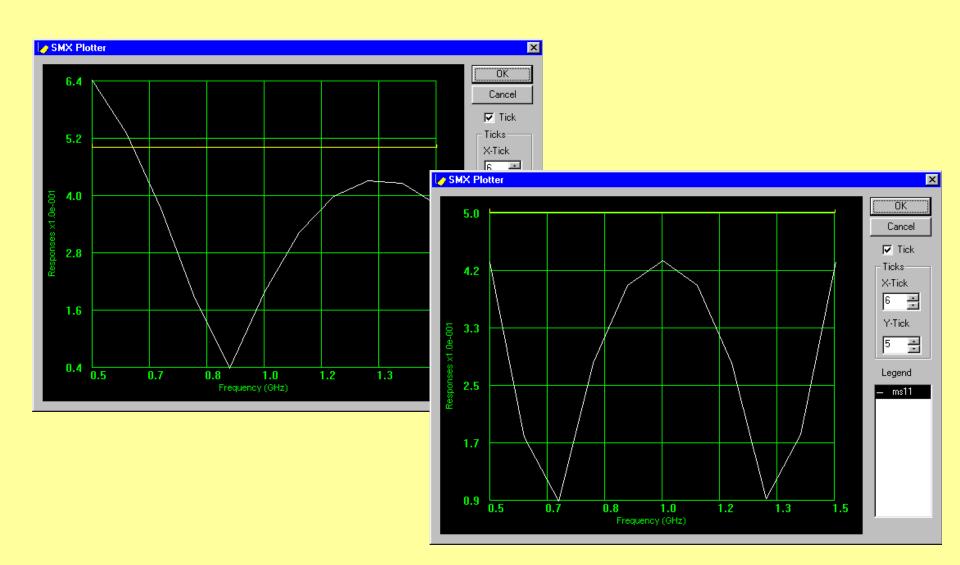


Object Oriented SMX System Example: Problem Setup Wizard

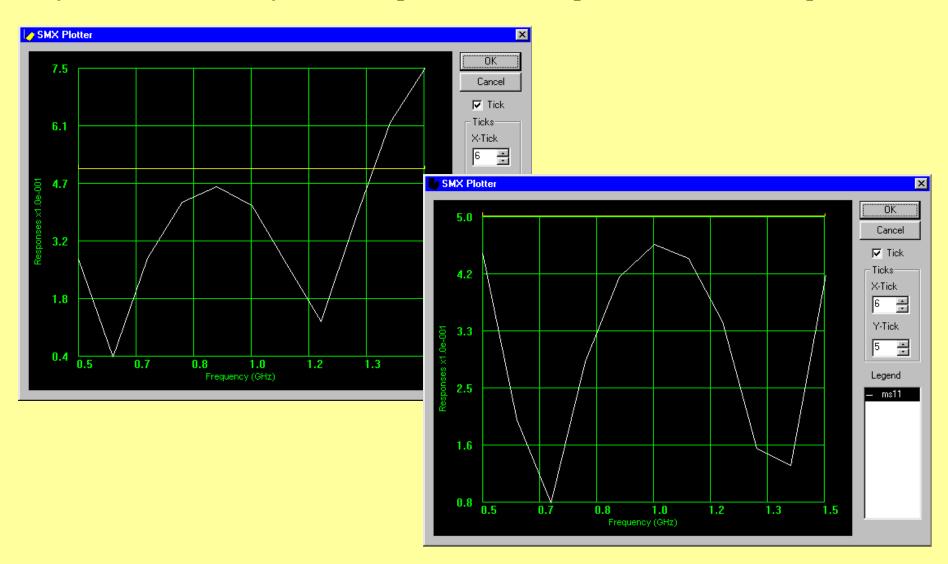
step 3: parameter setup



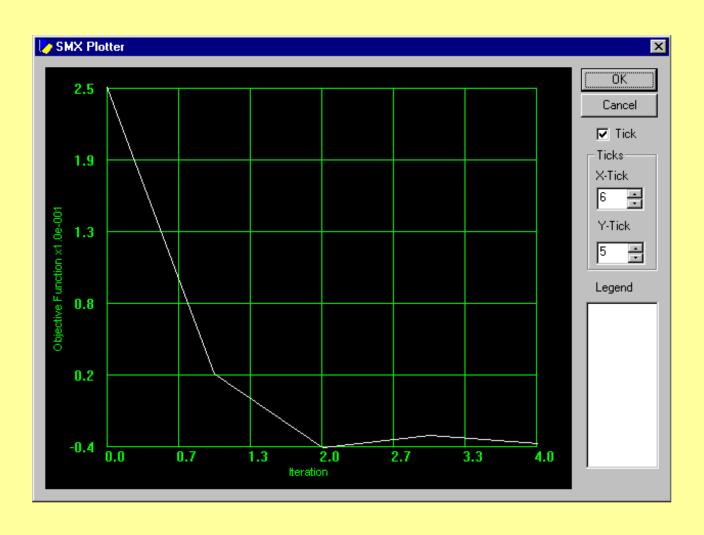
Object Oriented SMX System Example: Initial and Optimal Coarse Model Responses



Object Oriented SMX System Example: Initial and Optimal Fine Model Responses

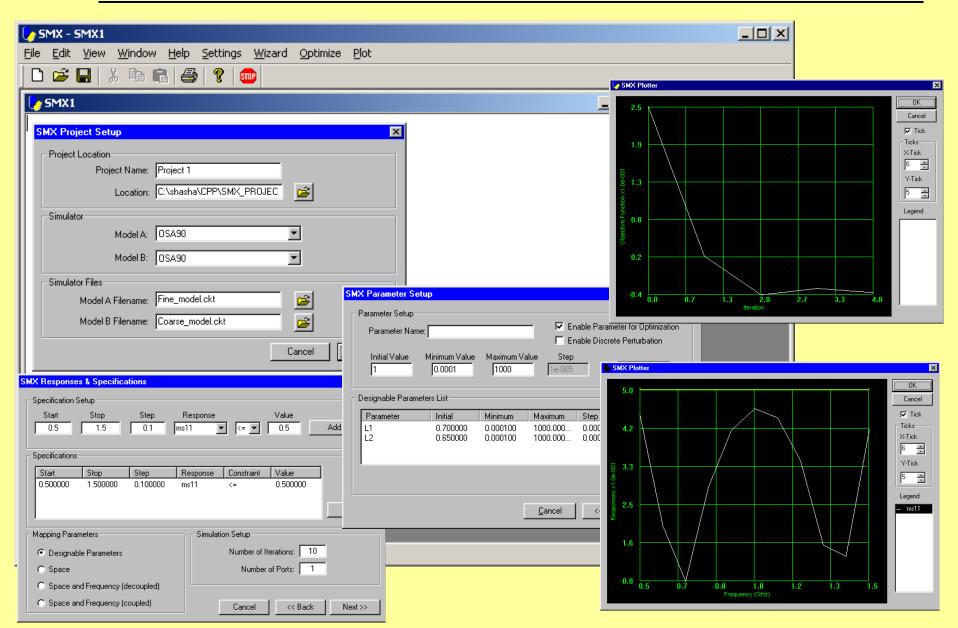


Object Oriented SMX System Example: Objective Function Value





Simulation Optimization Systems Research Laboratory McMaster University



Conclusions

we review Generalized Space Mapping (GSM) as a new engineering device modeling framework that exploits Frequency Space Mapping (FSM) and Multiple Space Mapping (MSM)

we review an innovative algorithm for EM optimization based on Space Mapping technology and Artificial Neural Networks

Neural Space Mapping (NSM) optimization exploits our SM-based neuromodeling techniques

a novel SM optimization algorithm based on surrogate models is presented

the surrogate model is a convex combination of a mapped coarse model and a linearized fine model

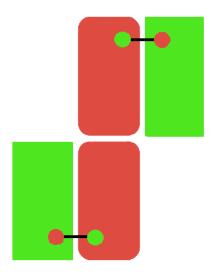
the state-of-the-art SMX engineering optimization system including Space Mapping technology is briefly reviewed

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First International Workshop on SURROGATE MODELLING AND SPACE MAPPING FOR ENGINEERING OPTIMIZATION

Technical University of Denmark Lyngby, Denmark November 16-18, 2000