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PRACTICAL LEAST  $p$ TH APPROXIMATION  
WITH EXTREMELY LARGE VALUES OF  $p$

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Abstract

Some computational experience relevant to computer-aided design using the ideas of least  $p$ th approximation with extremely large values of  $p$  is reported. Values of  $p$  up to 1,000,000,000,000 have been successfully employed in conjunction with efficient gradient minimization algorithms such as the Fletcher-Powell method and the recent method due to Fletcher.

1. INTRODUCTION

This paper describes some computational experience of the computer-aided design of actual circuits and systems using the ideas of least  $p$ th approximation with extremely large values of  $p$ . The obvious reason why large values of  $p$  are desirable is that the corresponding optimal approximations tend to become minimax (or Chebyshev) approximations. Thus, least  $p$ th approximation using the Fletcher-Powell method<sup>[1]</sup> and a new method due to Fletcher<sup>[2]</sup> and the values of  $p$  up to 1,000,000,000,000 have been tried on a number of test examples.

2. THE METHOD

We restrict ourselves here to a discussion of rather conventional discrete least  $p$ th approximation problems. Consider the minimization of

$$U(\phi) = \left( \sum_{i \in I} |e_i(\phi)|^p \right)^{\frac{1}{p}} > 0 \quad \text{for } 1 < p < \infty \quad (1)$$

where  $e_i(\phi)$ , in general, represents a weighted error or deviation between a complex specified function (desired response) and a complex approximating function (actual response) at some sample point  $i$  of a finite set  $I$ , and  $\phi$  represents the  $k$  variable parameters, i.e.,

$$\phi \triangleq [\phi_1 \quad \phi_2 \quad \dots \quad \phi_k]^T \quad (2)$$

Assuming that the  $e_i(\phi)$  for  $i \in I$  are continuous with continuous derivatives,

$$\nabla U(\phi) = \left( \sum_{i \in I} |e_i(\phi)|^p \right)^{\frac{1-p}{p}} \sum_{i \in I} |e_i(\phi)|^{p-2} \text{Re} \left\{ e_i^*(\phi) \nabla e_i(\phi) \right\} \dots (3)$$

where

$$\nabla \triangleq \left[ \frac{\partial}{\partial \phi_1} \quad \frac{\partial}{\partial \phi_2} \quad \dots \quad \frac{\partial}{\partial \phi_k} \right]^T \quad (4)$$

and  $*$  denotes the complex conjugate.

In an effort to alleviate the ill-conditioning resulting from the evaluation of  $|e_i(\phi)|^p$  for very large values of  $p$ , we let

$$M(\phi) \triangleq \max_{i \in I} |e_i(\phi)| \quad (5)$$

and rewrite (1) and (3), respectively, in the form

$$U(\phi) = M(\phi) \left( \sum_{i \in I} \left| \frac{e_i(\phi)}{M(\phi)} \right|^p \right)^{\frac{1}{p}} \quad (6)$$

and

$$\nabla U(\phi) = \left( \sum_{i \in I} \left| \frac{e_i(\phi)}{M(\phi)} \right|^p \right)^{\frac{1-p}{p}} \sum_{i \in I} \left| \frac{e_i(\phi)}{M(\phi)} \right|^{p-2} \text{Re} \left\{ \frac{e_i^*(\phi)}{M(\phi)} \nabla e_i(\phi) \right\} \dots (7)$$

noting that  $M(\phi) > 0$  and, again, that  $U(\phi) > 0$  and  $1 < p < \infty$ .

3. EXAMPLES

Space does not permit a full discussion of all the examples attempted, so we will briefly consider test problems for which the minimax solutions are known [3-6].

Consider the design of 10 $\Omega$  to 1 $\Omega$  transmission-line transformers for a relative bandwidth of 100%. Let  $e_i$  in (6) be  $\rho_i$ , the reflection coefficient, sampled uniformly at 11 points on the relative frequency interval [0.5, 1.5] GHz for 2-section designs and at {0.5, 0.6, 0.7, 0.77, 0.9, 1.0, 1.1, 1.23, 1.3, 1.4, 1.5} GHz for 3-section designs.

Appropriate gradient vectors with respect to length and characteristic impedance of transmission lines are calculated by the adjoint network method[7].

The progress of the two algorithms used on a CDC 6400 computer from indicated starting points is summarized in Tables 1 and 2 and Figures 1 to 5. For convenience we show results for different values of  $n$  where  $p=10^n$ . The figures are plots of  $M$  of (5) against  $N$ , the number of function evaluations at the beginning of an iteration. One function evaluation includes the evaluation of appropriate gradients.

#### 4. CONCLUSIONS

Our experience to-date indicates that, by and large, the smaller the value of  $p$ , the faster will  $M(p)$  be minimized, always assuming that the value of  $p$  is sufficiently large so that a specified  $M$  can be attained. No attempts at modifying the minimization methods to improve convergence for extremely large values of  $p$  nor a detailed study of other possible effects of numerical ill-conditioning have so far been carried out. But, if the success we have had using our present approach together with efficient gradient methods are widely repeatable, then far-reaching consequences are foreseen not only in nonlinear approximation, but in the closely related field of nonlinear programming[8]. Applications in the important area of filter design which can be readily carried out using least  $p$ th objectives[9], are also envisaged.

#### 5. ACKNOWLEDGEMENTS

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TABLE 1  
OPTIMIZATION OF A 2-SECTION 10:1 QUARTER-WAVE TRANSFORMER OVER 100 PERCENT BANDWIDTH WITH VARIABLE CHARACTERISTIC IMPEDANCES  $Z_1$  AND  $Z_2$

Fig.	Starting Point		$n$ where $p=10^n$	Number of Function Evaluations $N^*$	
	$Z_1$	$Z_2$		Fletcher [2]	Fletcher-Powell [1]
1	1.0	3.0	2	22	31
			3	28	49
			6	33	56
			9	33	56
			12	33	56
2	1.0	6.0	2	30	26
			3	58	50
			6	†	133
			9	†	172
			12	†	198
3	3.5	6.0	2	15	23
			3	†	41
			6	44	101
			9	102	118
			12	102	118
4	3.5	3.0	2	14	16
			3	19	56
			6	21	75
			9	21	85
			12	21	310

\*The number  $N$  listed are those required to bring  $M$  within 0.01 percent of the known optimum value, namely, 0.42857.

†Missing entries are due to parameters becoming negative - constraints were not imposed during optimization.

TABLE 2  
OPTIMIZATION OF A 3-SECTION 10:1 TRANSFORMER OVER 100 PERCENT BANDWIDTH WITH VARIABLE LENGTHS AND CHARACTERISTIC IMPEDANCES  
Starting Point:  $Z_1=1.5$ ,  $Z_2=3.0$ ,  $Z_3=6.0$ ,  $\ell_1/\ell_q=0.8$ ,  $\ell_2/\ell_q=1.2$ ,  $\ell_3/\ell_q=0.8$ , where  $\ell_q$  is the quarter-wavelength at center frequency.

$n$ where $p=10^n$	Number of Function Evaluations $N$ to reach the value of $M$ shown in brackets; the optimum value of $M$ is 0.19729	
	Fletcher [2]	Fletcher-Powell [1]
3	57 <del>36</del> (0.19734)	115 <del>43</del> (0.19734)
4	86 <del>104</del> (0.19730)	378 (0.19729)
6	418 (0.19729)	702 (0.19740)
9	634 <del>687</del> (0.19730)	661 (0.19740)
12	668 <del>684</del> (0.19736)	645 (0.19851)

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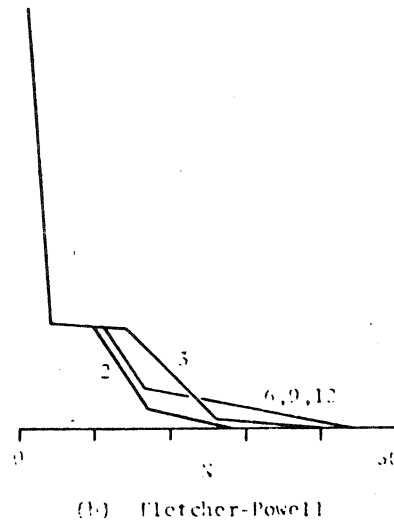
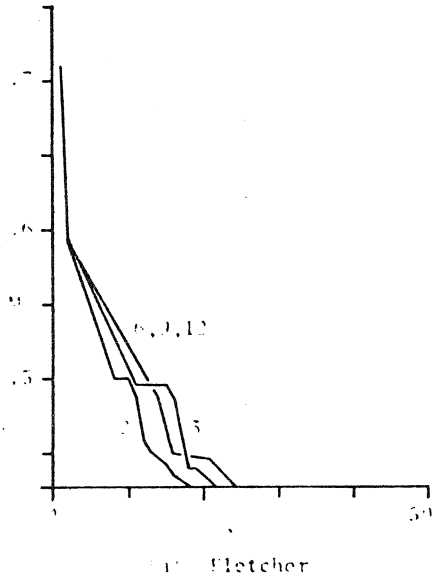


Fig. 1

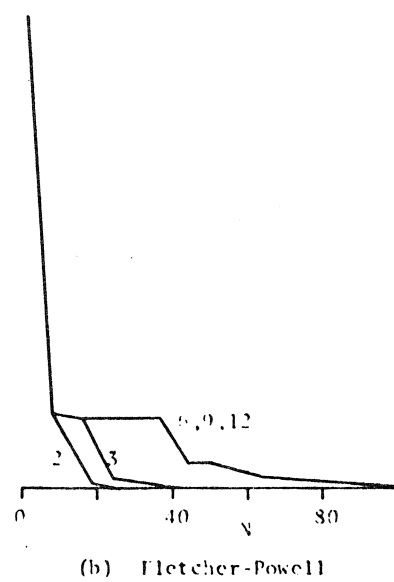
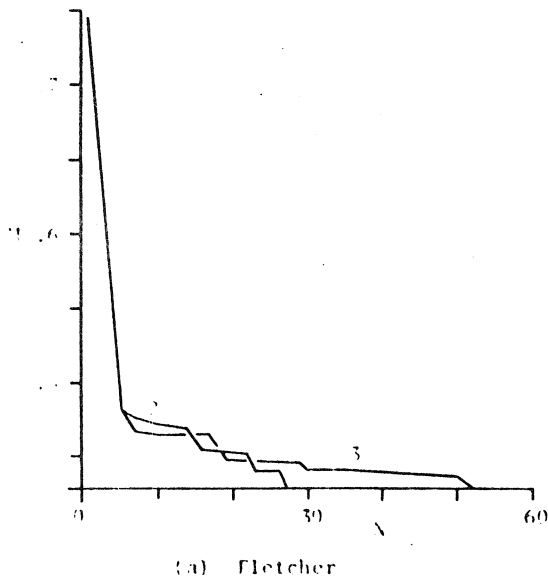
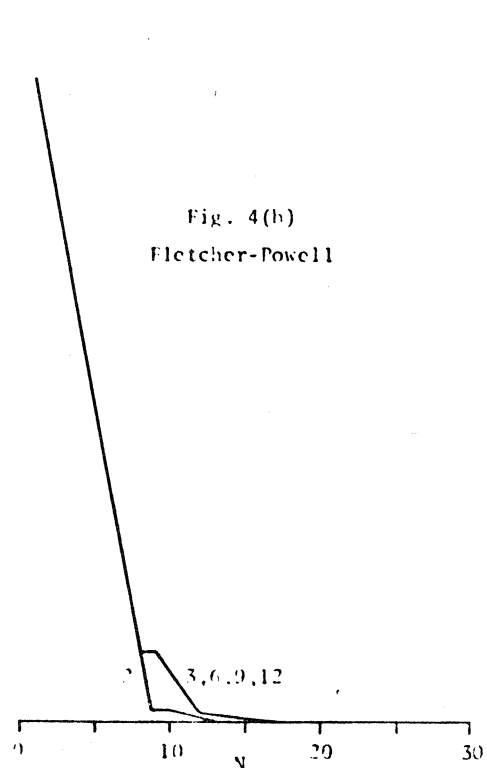
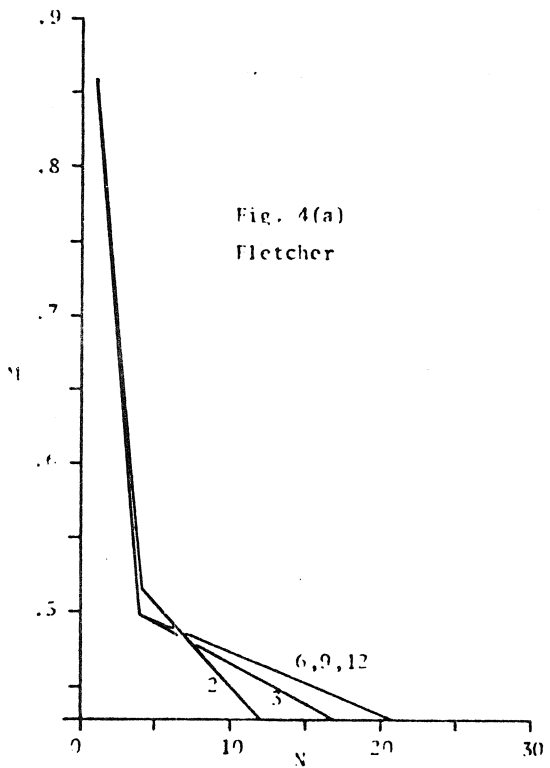
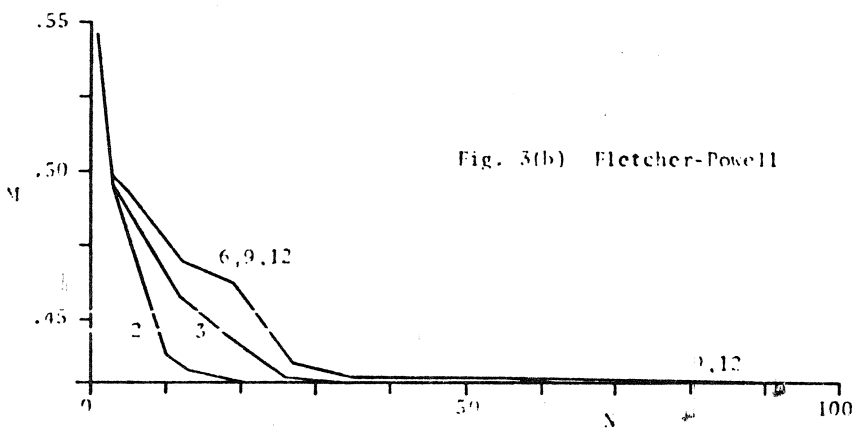
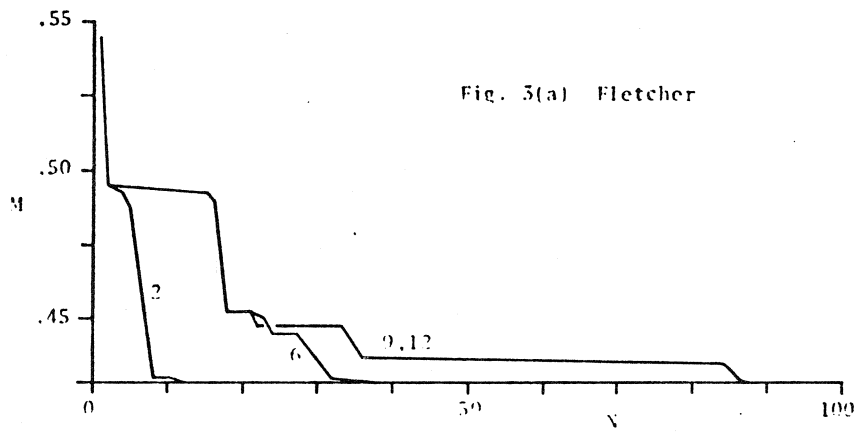


Fig. 2



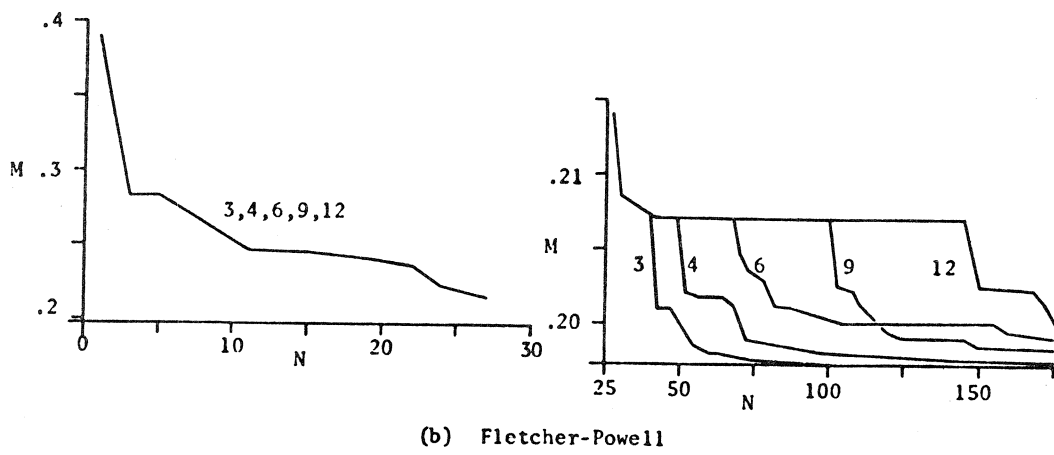
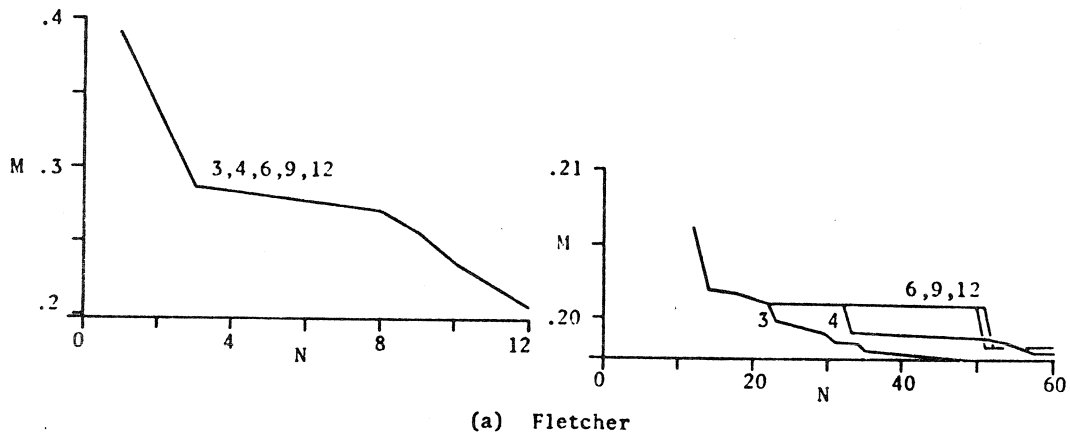


Fig. 5